

The Capacity of Noncoherent Continuous-Phase Frequency Shift Keying

Shi Cheng¹ Rohit Iyer Seshadri¹ Matthew C. Valenti¹ Don Torrieri²

¹Lane Department of Computer Science and Electrical Engineering
West Virginia University

²US Army Research Lab

March 14, 2007

Outline

- 1 Motivation
- 2 Noncoherent CPFSK
- 3 Capacity Analysis under Bandwidth Constraint
- 4 Application
- 5 Conclusion

Bandwidth Efficiency

$$x_i(t) = \frac{1}{\sqrt{T_S}} e^{j \frac{\pi(2k-(M-1))t}{T_S}}, \quad k = 0, 1, \dots, M-1$$

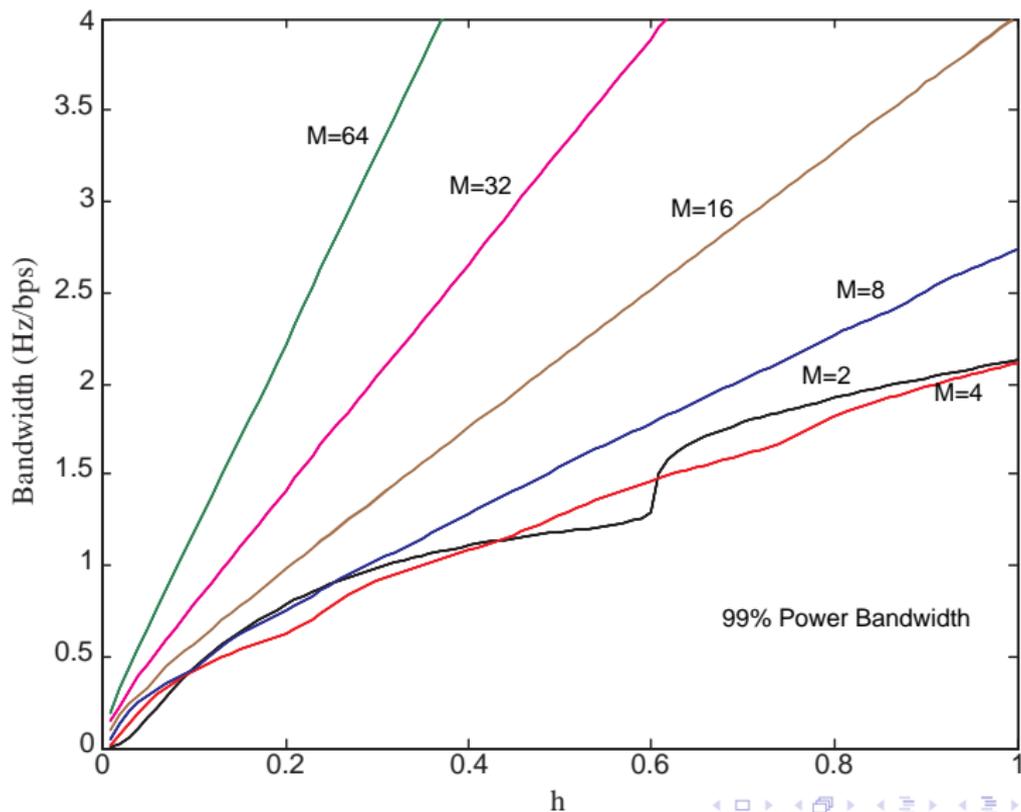
- Orthogonal FSK
 - Known to achieve Gaussian capacity when the number of tones M goes to infinity
 - Adjacent frequency tones are at least $1/T_S$ apart.
- Nonorthogonal FSK
 - Nonorthogonal FSK saves the bandwidth by using modulation index $h < 1$. Adjacent frequency tones are h/T_S apart.
 - Full response CPM with rectangular pulse shape, also called CPFSK
 - Partial response CPM has more compact bandwidth, but leads to complex signal processing

Bandwidth Efficiency

$$x_i(t) = \frac{1}{\sqrt{T_S}} e^{j\left(\frac{h\pi(2k-(M-1))t}{T_S} + \phi\right)}, \quad k = 0, 1, \dots, M-1$$

- Orthogonal FSK
 - Known to achieve Gaussian capacity when the number of tones M goes to infinity
 - Adjacent frequency tones are at least $1/T_S$ apart.
- Nonorthogonal FSK
 - Nonorthogonal FSK saves the bandwidth by using **modulation index** $h < 1$. Adjacent frequency tones are h/T_S apart.
 - Full response CPM with rectangular pulse shape, also called **CPFSK**
 - Partial response CPM has more compact bandwidth, but leads to complex signal processing

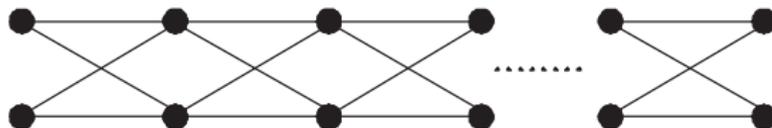
Bandwidth of CPFSK



CPFSK Detection

- **Coherent** Detection

- Decoding through trellis
- h needs to be rational
- Phase synchronization

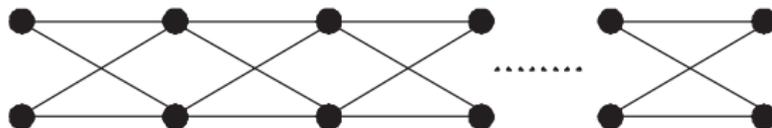


- **Noncoherent** Detection

- Symbol by symbol noncoherent detection

CPFSK Detection

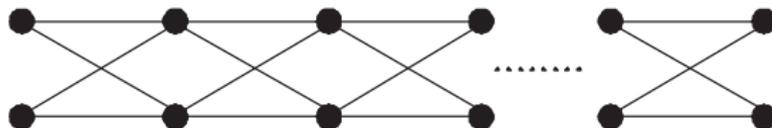
- **Coherent** Detection
 - Decoding through trellis
 - h needs to be rational
 - Phase synchronization



- **Noncoherent** Detection
 - Symbol by symbol noncoherent detection

CPFSK Detection

- **Coherent** Detection
 - Decoding through trellis
 - h needs to be rational
 - Phase synchronization



- **Noncoherent** Detection
 - Symbol by symbol noncoherent detection

Noncoherent CPFSK Discrete Time Model

$$\mathbf{y} = ae^{j\theta} \sqrt{\mathcal{E}_s} \mathbf{x} + \mathbf{n}$$

- \mathbf{n} is colored noise, with $E(\mathbf{nn}^H) = N_0 \mathbf{K}$
- \mathbf{x} is chosen from columns of $\mathbf{K} = [\mathbf{k}_0, \mathbf{k}_1, \dots, \mathbf{k}_{M-1}]$
- $p(\mathbf{y} | \mathbf{x} = \mathbf{k}_\nu) \propto I_0 \left(2 \frac{a\sqrt{\mathcal{E}_s}}{N_0} |y_\nu| \right)$

A Binary Example: Minimum Shift Keying (MSK)

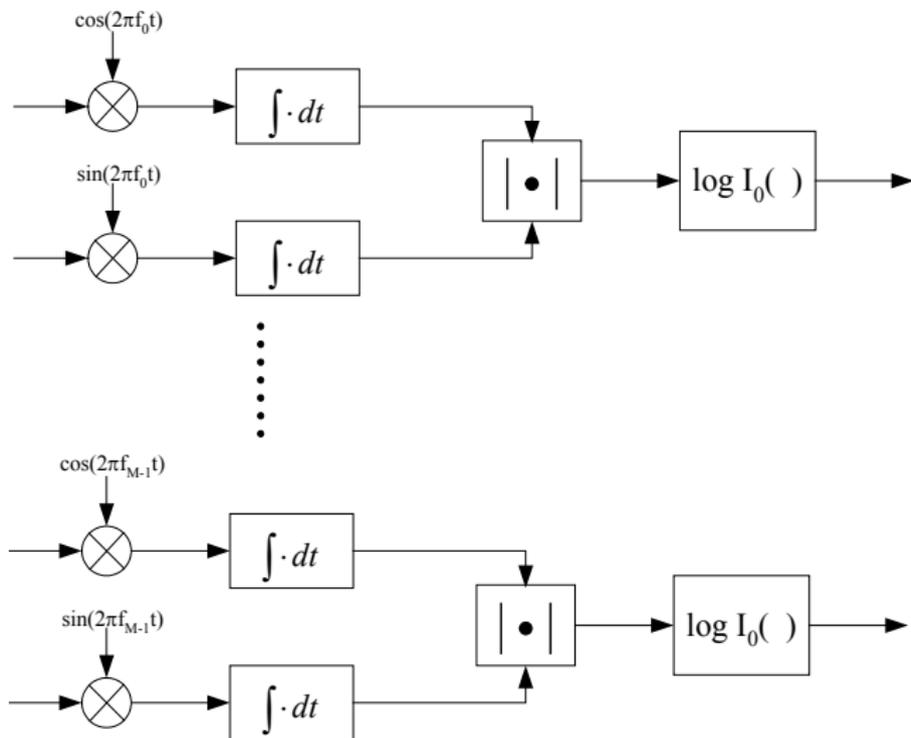
- MSK: $M = 2, h = 1/2$
- The normalized correlation matrix for \mathbf{n} is

$$\mathbf{K} = \begin{bmatrix} 1 & -0.6366j \\ 0.6366j & 1 \end{bmatrix}$$

- The modulator selects the columns of \mathbf{K} , depending on the M-ary input d .

	$d = 0$	$d = 1$
$\mathbf{x} =$	$\begin{bmatrix} 1 \\ -0.6366j \end{bmatrix}$	$\begin{bmatrix} 0.6366j \\ 1 \end{bmatrix}$

Noncoherent CPFSK Detector Diagram



Capacity Calculation

- Instantaneous capacity

$$i(\mathbf{x}; \mathbf{y}) = \log M - \log \frac{\sum_{\mathbf{x}' \in \mathcal{S}} p(\mathbf{y}|\mathbf{x}')}{p(\mathbf{y}|\mathbf{x})}.$$

- Capacity

$$I(\mathbf{x}; \mathbf{y}) = \log M - E_{\mathbf{x}} \left[\log \frac{\sum_{\mathbf{x}' \in \mathcal{S}} p(\mathbf{y}|\mathbf{x}')}{p(\mathbf{y}|\mathbf{x})} \right].$$

- Monte Carlo simulation

Capacity Calculation

- Instantaneous capacity

$$i(\mathbf{x}; \mathbf{y}) = \log M - \log \frac{\sum_{\mathbf{x}' \in \mathcal{S}} p(\mathbf{y}|\mathbf{x}')}{p(\mathbf{y}|\mathbf{x})}.$$

- Capacity

$$I(\mathbf{x}; \mathbf{y}) = \log M - E_{\mathbf{x}} \left[\log \frac{\sum_{\mathbf{x}' \in \mathcal{S}} p(\mathbf{y}|\mathbf{x}')}{p(\mathbf{y}|\mathbf{x})} \right].$$

- Monte Carlo simulation

Capacity Calculation

- Instantaneous capacity

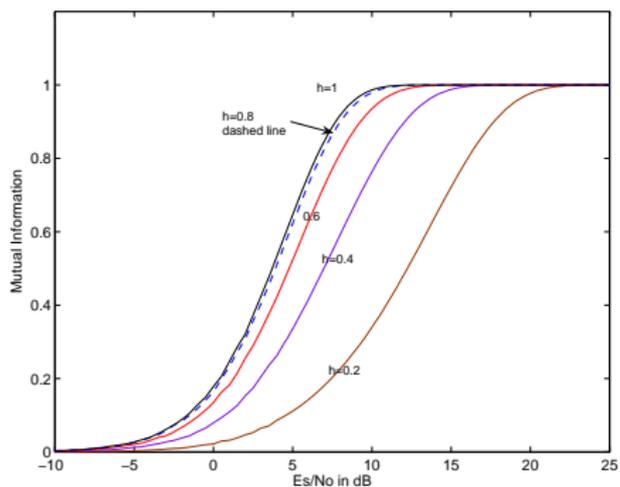
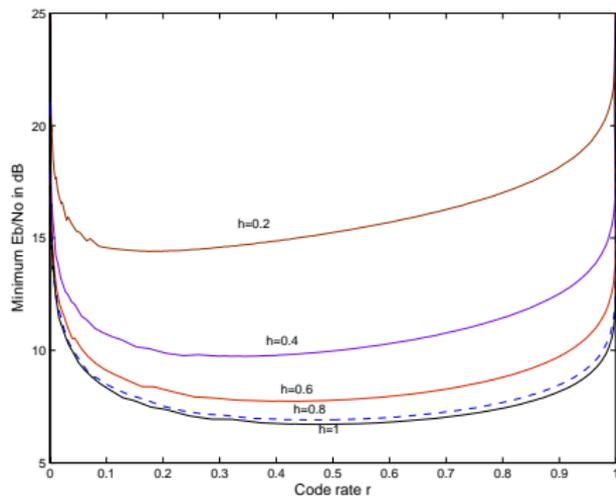
$$i(\mathbf{x}; \mathbf{y}) = \log M - \log \frac{\sum_{\mathbf{x}' \in \mathcal{S}} p(\mathbf{y}|\mathbf{x}')}{p(\mathbf{y}|\mathbf{x})}.$$

- Capacity

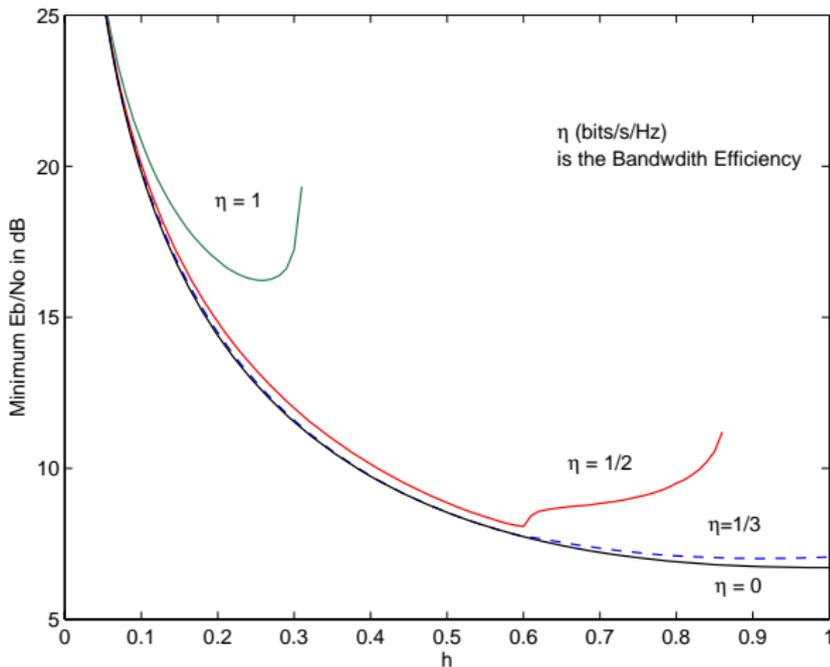
$$I(\mathbf{x}; \mathbf{y}) = \log M - E_{\mathbf{x}} \left[\log \frac{\sum_{\mathbf{x}' \in \mathcal{S}} p(\mathbf{y}|\mathbf{x}')}{p(\mathbf{y}|\mathbf{x})} \right].$$

- Monte Carlo simulation

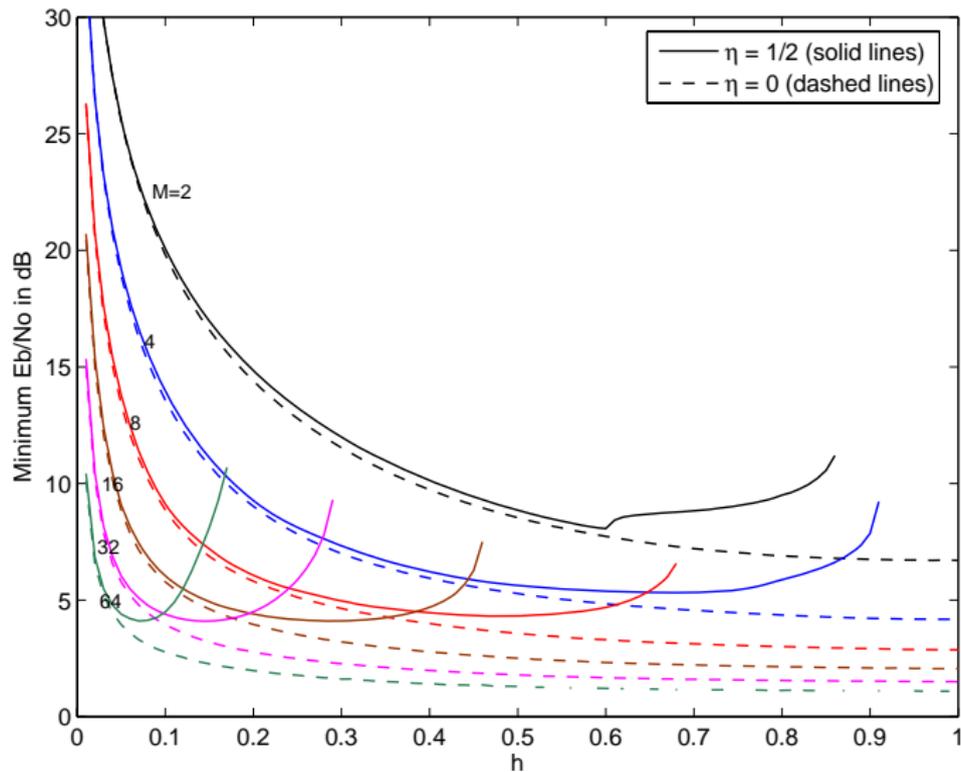
Binary Noncoherent CPFSK Capacities in AWGN

(a) channel capacity versus \mathcal{E}_S/N_0 (b) minimum \mathcal{E}_b/N_0 versus coding rate

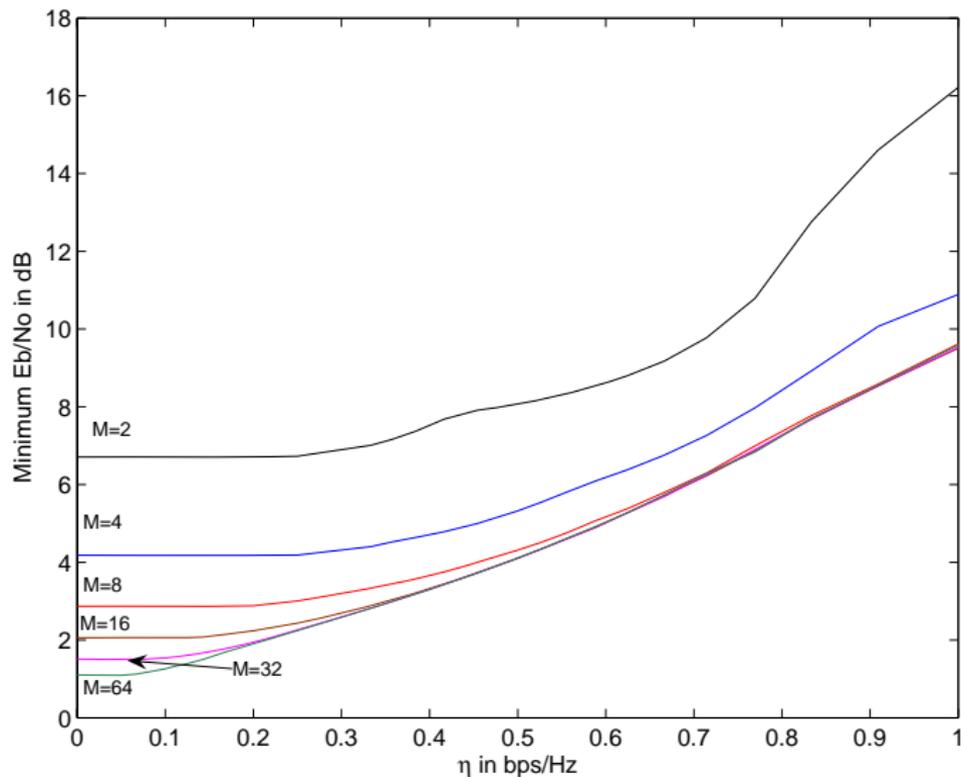
Binary Noncoherent CPFSK Capacities in AWGN under Bandwidth Constraint



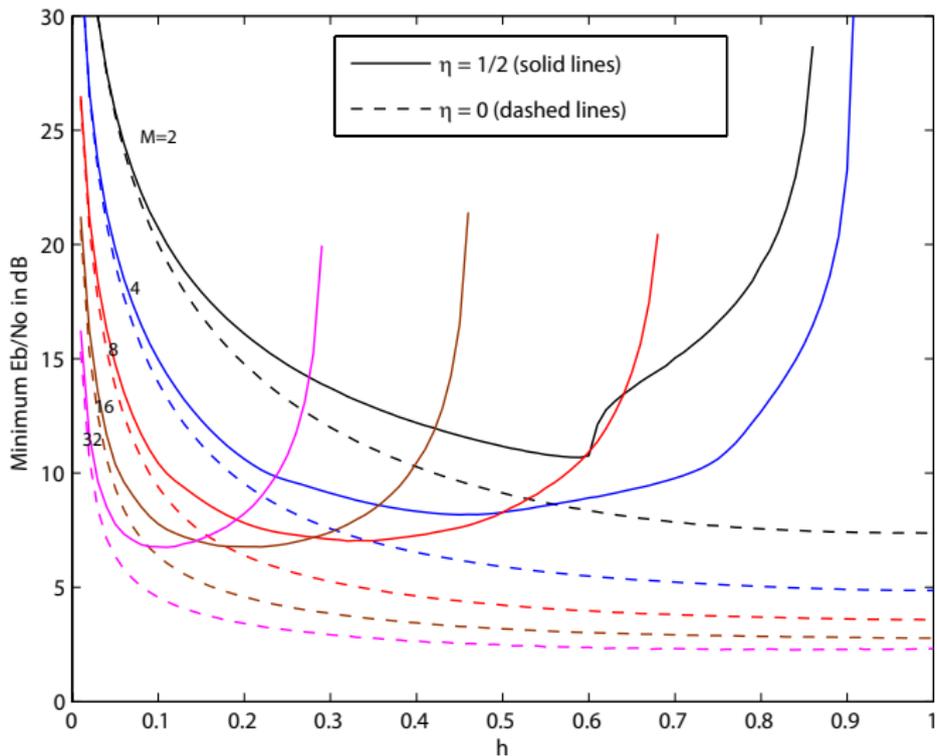
Noncoherent CPFSK Capacities in AWGN Channel



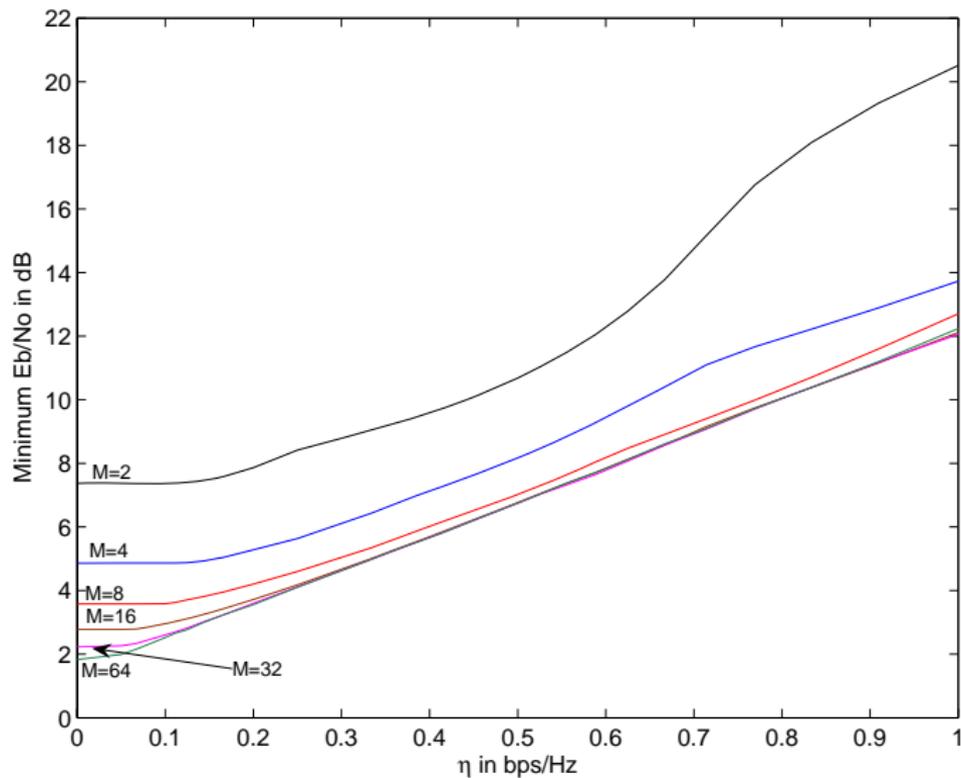
Noncoherent CPFSK Capacities in AWGN Channel



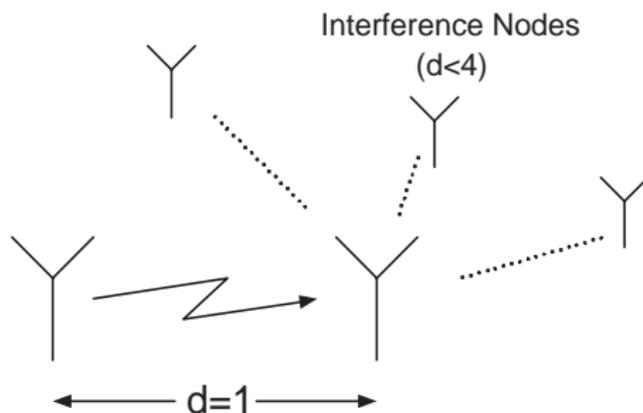
Noncoherent CPFSK Capacities in Rayleigh Channel



Noncoherent CPFSK Capacities in Rayleigh Channel



FH Network Scenario



- Multiple access interference
- Tradeoff on the bandwidth of each sub-channel and the possibility of interference
- Equal received average SNR, Rayleigh fading, Interference nodes subject to log normal shadowing $\sigma = 8dB$

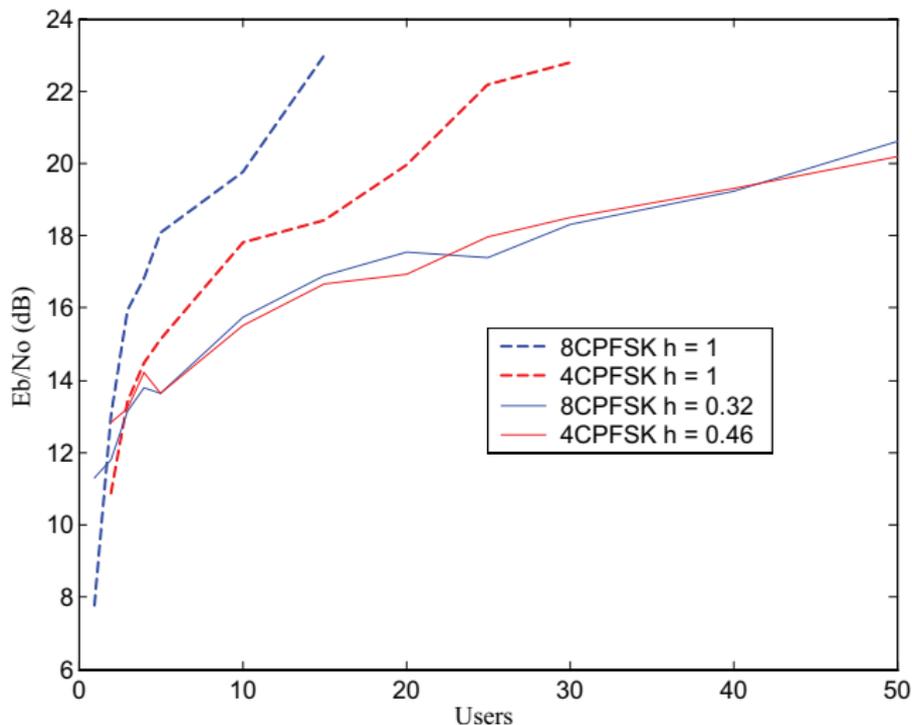
FH Network Scenario

- Total bandwidth $W = 2000/T_u$

h	M	Coding rate	Number of channels
1	4	2048/6144	315
1	8	2048/6144	244
0.46	4	2048/3456	1000
0.32	8	2048/3840	1000

- All users use the same modulation and coding strategy
- Channel estimator using EM algorithm, the same as the one of orthogonal FSK

Noncoherent CPFSK FH Network against MAI



- SNR collected at $BER = 10^{-4}$
- UMTS turbo code
- 32 hops

Conclusion

- This paper outlines a methodology for finding the coded modulation (CM) capacity of CPFSK with noncoherent detection.
- For a specific number of tones M and spectral efficiency η , the minimum required \mathcal{E}_b/N_0 can be optimized over the modulation index h and coding rate r .
- For a fixed spectral efficiency, it is better to use a higher order M and smaller h . However, when the spectral efficiency is sufficiently high, the benefit is small by using $M > 8$.

Thank you