Physical-layer Network Coding using FSK Modulation under Frequency Offset

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Introduction

- Physical-layer network coding (PNC) is a capacity increasing protocol which exploits simultaneous transmission by multiple nodes in the same time and band.
- This work applies PNC to the two-way relay channel under constant envelope, noncoherent, orthogonal modulation and error correction coding.
- The specific focus is to investigate the error-rate penalty of carrier frequency offset on the MAC transmission phase.
- Certain contemporary radio platforms such as USRP software radios contain high-tolerance oscillators which generate imperfect carrier frequencies.

Simulated Error Rate Performance - no Error Correction Code





System Model

- Network Topology: Two-way Relay Channel
- Protocol: Physical-Layer Network Coding (MAC phase Sources to Relay)
- Channel Model: Frequency-flat slowly-varying Rayleigh fading
- Modulation: Orthogonal Binary Frequency-shift Keying
- Channel code: Turbo coded and Uncoded



Blue, dashed lines: detector does not compensate for offset. Black, solid lines: detector compensates for offset. $d_1 = 0$

- Detection rule that does not incorporate offset reaches a minimum error rate and then degrades in performance.
- Rule incorporating offset outperforms the rule which does not, however, the error rate encounters an error floor.
- SNR required to reach error rate 10^{-2} :

No Offset Compensation		Offset Incorpo	Offset Incorporated	
Frequency offset (d _i) SNR (dB)		Frequency offset	Frequency offset SNR	
0.04	N/A	0.04	32	
0.02	32	0.02	27	
0	26	0	26	

Simulation Results - Turbo Channel Coding



- We show that oscillator offset may be modeled as a linear, multiplicative effect.
 The channel statistics at the output of the relay demodulator are
 $r = h_1 O_1 s_1 + h_2 O_2 s_2 + n$
- \blacktriangleright The matrices O_1 and O_2 quantifying offset are defined as

$$O_{i} = \begin{bmatrix} O_{i}[0, 1] & O_{i}[1, 1] \\ O_{i}[0, 2] & O_{i}[1, 2] \end{bmatrix} \ i \in \{1, 2\}$$

▶ The elements of O_1 and O_2 are

$$O_{i}[b_{i},m] = \begin{bmatrix} \frac{\sin A_{i,m}}{A_{i,m}} - j \frac{\cos A_{i,m} - 1}{A_{i,m}} \end{bmatrix}$$

where b_i denotes the bit transmitted by source i and m denotes the demodulator output corresponding to FSK tone m, and the term $A_{i,m}$ is $A_{i,m} = 2\pi\{[b_i - (m-1)] + d_i/\Delta_f\}$

where Δ_f is the spacing between FSK tones, and d_i is the frequency offset.

Relay Detection Rule

Goal: derive optimal detection rule for network-coded bit at relay defined by the log-likelihood ratio of the network-coded bit Blue, dashed lines: detector does not compensate for offset. Black, solid lines: detector compensates for offset. $d_1 = 0$ Channel code rate r=4500/6500

- \blacktriangleright Turbo Channel lowers error floor exhibited by rule incorporating offset beneath an error rate of 10^{-5}
- SNR required to reach error rate 10^{-3} :

$$\Lambda(\mathbf{b}) = \log \frac{\mathsf{P}(\mathbf{b} = 1)}{\mathsf{P}(\mathbf{b} = 0)} = \log \frac{\mathsf{P}(\mathbf{b}_1 \oplus \mathbf{b}_2 = 1)}{\mathsf{P}(\mathbf{b}_1 \oplus \mathbf{b}_2 = 0)}$$

- The received signal at the relay may be written in vector form as $\mathbf{r} = \begin{bmatrix} h_1 O_1[\mathbf{b}_1, \mathbf{1}] + h_2 O_2[\mathbf{b}_2, \mathbf{1}] + \mathbf{n}_1 \\ h_1 O_1[\mathbf{b}_1, \mathbf{2}] + h_2 O_2[\mathbf{b}_2, \mathbf{2}] + \mathbf{n}_2 \end{bmatrix}$
- The demodulator outputs are the sum of zero-mean i.i.d. Gaussian RV's.
 The outputs are thus complex jointly Gaussian RV's completely described by the covariance matrix K.

$$\label{eq:prod} p(r|S_i) = \frac{1}{(2\pi)^2|\mathsf{K}|} \exp\left(-r^\mathsf{H}\mathsf{K}^{-1}r\right)$$
 where S_i is the symbol pair transmitted by the end nodes.

No Offset Compensation		Offset Incorpor	Offset Incorporated		
Frequency offset (d _i	SNR (dB)	Frequency offset	SNR		
0.06	28	0.06	25		
0.04	24.5	0.04	24.5		
0	24	0	24		

Conclusion

- Result: Noncoherent demodulator and detector for the MAC phase of the TWRC which compensates for carrier frequency offset.
- Detector exhibits error rates below 10⁻⁵ for offsets on the order of a few hundredths of a frequency spacing when paired with a Turbo channel code.
 Uncoded error rates on the order of 10⁻². Further analysis is required to discover the cause of and eliminate the error floor.

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