

# Power Allocation for Distributed BLUE Estimation with Full and Limited Feedback of CSI

**Mohammad Fanaei**, Matthew C. Valenti, and  
Natalia A. Schmid

Department of Computer Science and Electrical Engineering  
West Virginia University, Morgantown, WV, U.S.A.

Military Communications Conference (Milcom'13)  
Track 1: Waveforms and Signal Processing

November 18, 2013

- 1 Introduction and Problem Statement
- 2 System Model Description
- 3 Adaptive Optimal Power-Allocation Scheme
- 4 Limited Feedback for Adaptive Power Allocation
- 5 Numerical Results
- 6 Conclusions

# Outline

- 1 Introduction and Problem Statement
- 2 System Model Description
- 3 Adaptive Optimal Power-Allocation Scheme
- 4 Limited Feedback for Adaptive Power Allocation
- 5 Numerical Results
- 6 Conclusions

# What Is Distributed Estimation in WSNs?

## Spatially distributed sensors

- Observe their surrounding environment.
- Process their local observations.
- Send their processed data to a fusion center (FC).
- FC performs the ultimate global estimation.

# What Is Distributed Estimation in WSNs?

Spatially distributed sensors

- Observe their surrounding environment.
- Process their local observations.
- Send their processed data to a fusion center (FC).
- FC performs the ultimate global estimation.

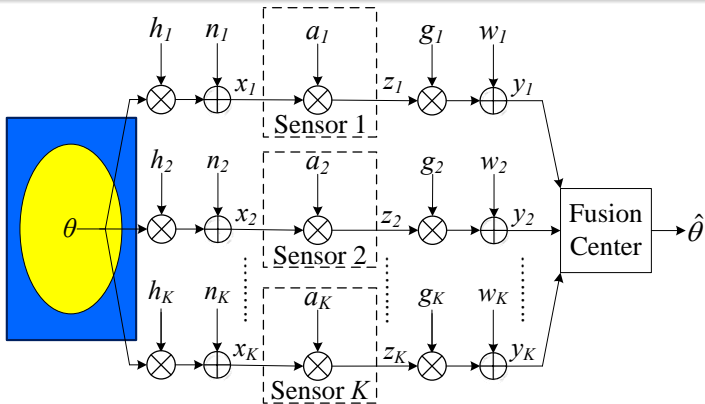
## Goal of This Paper

Adaptive power allocation to local sensors for *reliable* estimation of an unknown random parameter at the fusion center of a WSN from spatially distributed linear noisy observations by local sensors and communicated through parallel fading channels.

# Outline

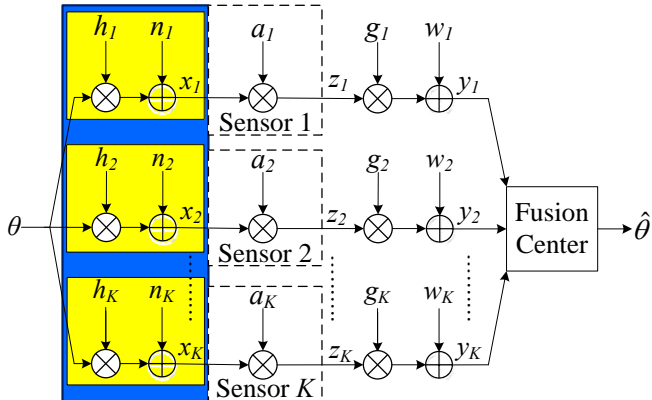
- 1 Introduction and Problem Statement
- 2 System Model Description**
- 3 Adaptive Optimal Power-Allocation Scheme
- 4 Limited Feedback for Adaptive Power Allocation
- 5 Numerical Results
- 6 Conclusions

# System Model Description



Unknown random parameter  $\theta$  with zero mean and unit power, otherwise unknown.

## System Model Description



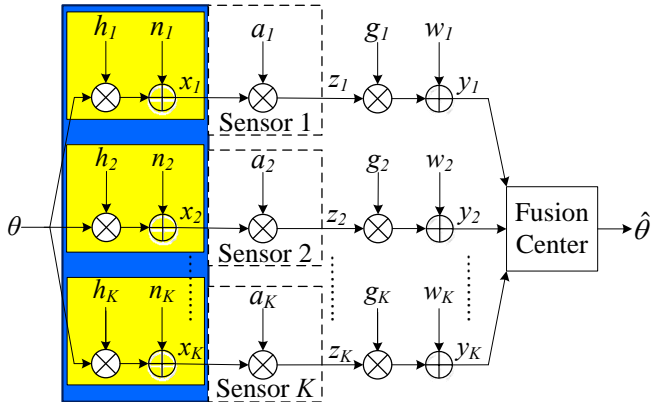
- Local linear noisy observation at each sensor

$$x_i = h_i \theta + n_i$$

- $h_i$ : Fixed local observation gain known at the sensor and FC.



## System Model Description

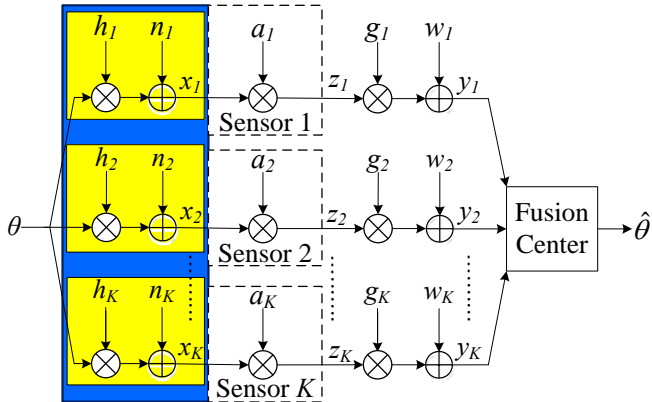


- Local linear noisy observation at each sensor

$$x_i = h_i\theta + n_i$$

- $n_i$ : Spatially i.i.d. additive observation noise with zero mean and known variance  $\sigma_o^2$ , otherwise unknown.

## System Model Description

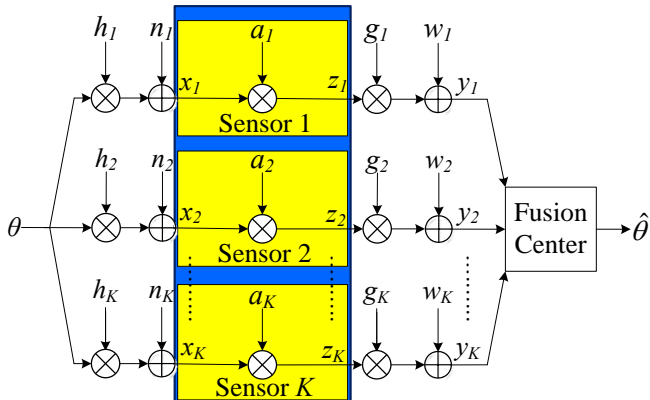


- Local linear noisy observation at each sensor

$$x_i = h_i \theta + n_i$$

- Observation SNR at sensor  $i$ :  $\beta_i \stackrel{\text{def}}{=} \frac{|h_i|^2}{\sigma_n^2}$

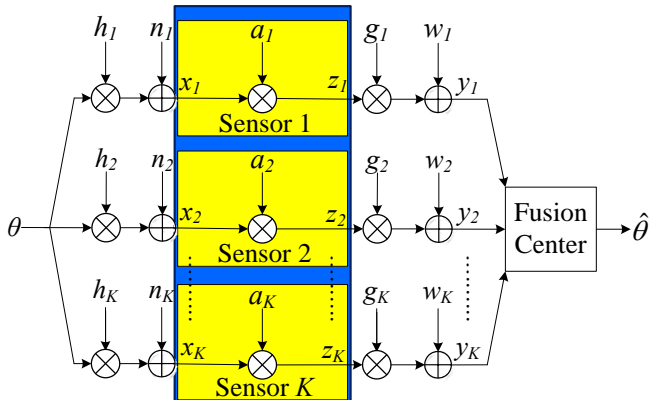
# System Model Description



Amplify-and-forward (AF) local processing scheme

- No inter-sensor communication and/or collaboration.
- Each sensor acts as a relay and transmits amplified version of its raw noisy analog local observation to the fusion center.

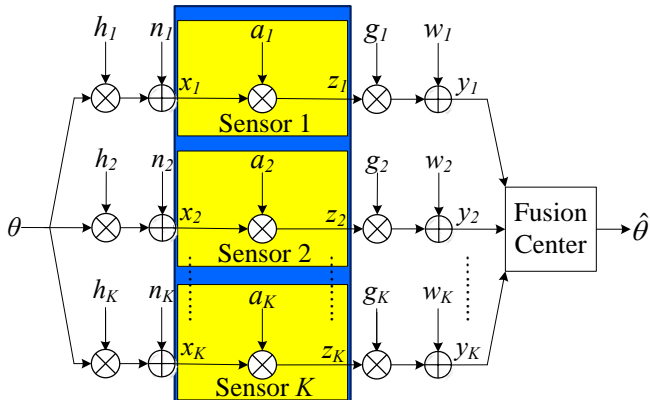
## System Model Description



Amplify-and-forward (AF) local processing scheme

$$z_i = a_i x_i = a_i h_i \theta + a_i n_i$$

## System Model Description



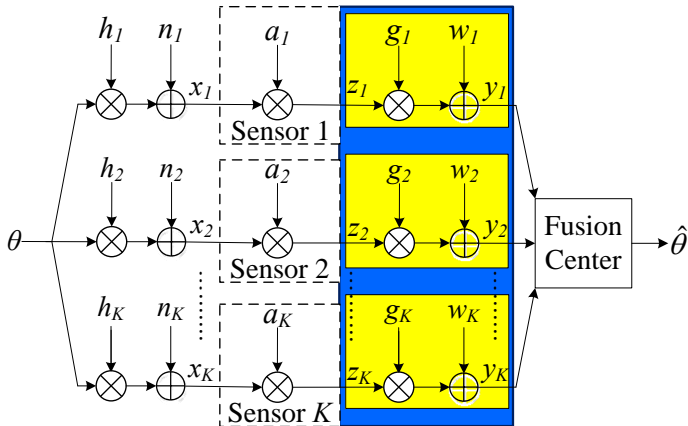
Amplify-and-forward (AF) local processing scheme

$$z_i = a_i x_i = a_i h_i \theta + a_i n_i$$

- Instantaneous transmit power

$$P_i = a_i^2 \left( |h_i|^2 + \sigma_0^2 \right) = a_i^2 \sigma_0^2 (1 + \beta_i)$$

## System Model Description

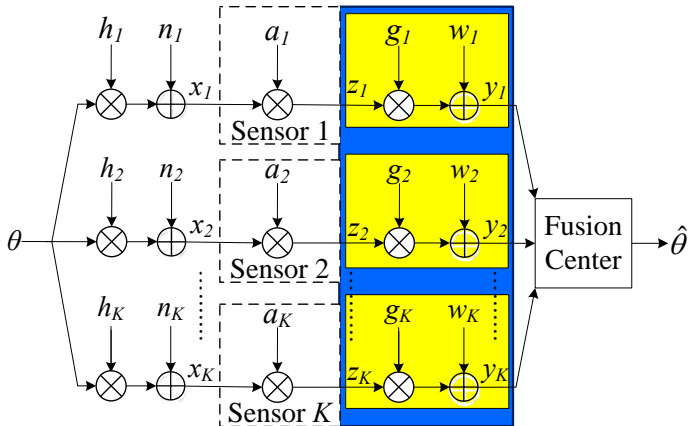


Orthogonal coherent channels corrupted by independent fading and additive Gaussian noise:

$$y_i = g_i z_i + w_i \quad w_i \sim \mathcal{N}(0, \sigma_c^2)$$

Channel gains are completely *known* at the fusion center.

## System Model Description

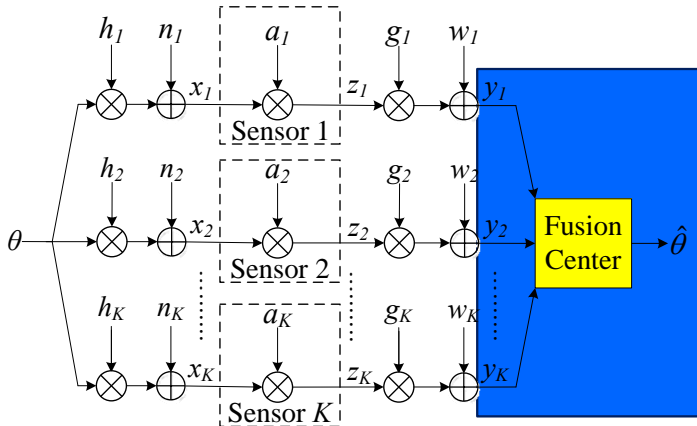


Orthogonal coherent channels corrupted by independent fading and additive Gaussian noise:

$$y_i = g_i z_i + w_i \quad w_i \sim \mathcal{N}(0, \sigma_c^2)$$

- Channel SNR at sensor  $i$ :  $\gamma_i \stackrel{\text{def}}{=} \frac{|g_i|^2}{\sigma_c^2}$

## System Model Description



The FC finds the *best linear unbiased estimator* (BLUE) for  $\theta$ .



# BLUE Estimation of $\theta$ at the FC

- Given  $\mathbf{a} \triangleq [a_1, a_2, \dots, a_K]^T$  and a realization of the fading channel gains  $\mathbf{g} \triangleq [g_1, g_2, \dots, g_K]^T$ , the BLUE estimator of  $\theta$  at the FC is

$$\hat{\theta} = \left( \sum_{i=1}^K \frac{h_i^2 a_i^2 g_i^2}{a_i^2 g_i^2 \sigma_o^2 + \sigma_c^2} \right)^{-1} \sum_{i=1}^K \frac{h_i a_i g_i y_i}{a_i^2 g_i^2 \sigma_o^2 + \sigma_c^2}$$

- The estimator variance is

$$\begin{aligned} \text{Var}(\hat{\theta} | \mathbf{a}, \mathbf{g}) &= \left( \sum_{i=1}^K \frac{h_i^2 a_i^2 g_i^2}{a_i^2 g_i^2 \sigma_o^2 + \sigma_c^2} \right)^{-1} \\ &= \left( \sum_{i=1}^K \frac{\beta_i \gamma_i a_i^2 \sigma_o^2}{1 + \gamma_i a_i^2 \sigma_o^2} \right)^{-1} \end{aligned}$$

# Outline

- 1 Introduction and Problem Statement
- 2 System Model Description
- 3 Adaptive Optimal Power-Allocation Scheme**
- 4 Limited Feedback for Adaptive Power Allocation
- 5 Numerical Results
- 6 Conclusions

# Problem Formulation

## Problem Statement for Our Optimal Power-Allocation Scheme

Find the optimal local amplification gains or equivalently, the optimal power-allocation scheme that minimizes the  $L^2$ -norm of the vector of local transmit powers  $\mathbf{P} \triangleq [P_1, P_2, \dots, P_K]^T$ , given a maximum estimation distortion as defined by the BLUE estimator variance.

$$\begin{aligned} & \text{minimize}_{\{P_i\}_{i=1}^K} \left( \sum_{i=1}^K P_i^2 \right)^{\frac{1}{2}} \\ & \text{subject to } \text{Var} \left( \hat{\theta} | \mathbf{a}, \mathbf{g} \right) \leq D_0 \end{aligned}$$

# Problem Formulation

## Problem Statement for Our Optimal Power-Allocation Scheme

Find the optimal local amplification gains or equivalently, the optimal power-allocation scheme that minimizes the  $L^2$ -norm of the vector of local transmit powers  $\mathbf{P} \triangleq [P_1, P_2, \dots, P_K]^T$ , given a maximum estimation distortion as defined by the BLUE estimator variance.

$$\begin{aligned} & \underset{\{a_i\}_{i=1}^K}{\text{minimize}} && \sum_{i=1}^K [a_i^2 \sigma_0^2 (1 + \beta_i)]^2 \\ & \text{subject to} && \sum_{i=1}^K \frac{\beta_i \gamma_i a_i^2 \sigma_0^2}{1 + \gamma_i a_i^2 \sigma_0^2} \geq \frac{1}{D_0} \end{aligned}$$

# Optimal Local Amplification Gains

The unique solution for the given constrained *convex* optimization problem could be found as follows:

$$a_i^2 = \begin{cases} \frac{1}{\gamma_i \sigma_0^2} \left( \frac{\sqrt[3]{\frac{\lambda_0}{\beta_i \delta_i^2 T_i}}}{1 - \frac{2}{3}} \sqrt[3]{\frac{\beta_i \delta_i^2}{\lambda_0 T_i^2}} - 1 \right), & i \leq K_1 \\ 0, & i > K_1 \end{cases}$$

$$T_i = 1 + \sqrt{1 + \frac{8\beta_i \delta_i^2}{27\lambda_0}} \quad \delta_i \stackrel{\text{def}}{=} \frac{1 + \beta_i}{\beta_i \gamma_i} \quad \beta_i \stackrel{\text{def}}{=} \frac{|h_i|^2}{\sigma_0^2} \quad \text{and} \quad \gamma_i \stackrel{\text{def}}{=} \frac{|g_i|^2}{\sigma_c^2}$$

- The sensors are sorted so that  $\delta_1 \leq \delta_2 \leq \dots \leq \delta_K$ .
  - The observation SNR  $\beta_i$  or channel SNR  $\gamma_i$  decreases.
- *Only* the first  $K_1$  sensors with the least values of  $\delta_i$  will have a positive value for  $a_i$ , and  $a_i = 0$  for all  $i > K_1$ .

# Optimal Local Amplification Gains

The unique solution for the given constrained *convex* optimization problem could be found as follows:

$$a_i^2 = \begin{cases} \frac{1}{\gamma_i \sigma_0^2} \left( \frac{\sqrt[3]{\frac{\lambda_0}{\beta_i \delta_i^2 T_i}}}{1 - \frac{2}{3} \sqrt[3]{\frac{\beta_i \delta_i^2}{\lambda_0 T_i^2}}} - 1 \right), & i \leq K_1 \\ 0, & i > K_1 \end{cases}$$

$$T_i = 1 + \sqrt{1 + \frac{8\beta_i \delta_i^2}{27\lambda_0}} \quad \delta_i \stackrel{\text{def}}{=} \frac{1 + \beta_i}{\beta_i \gamma_i} \quad \beta_i \stackrel{\text{def}}{=} \frac{|h_i|^2}{\sigma_0^2} \quad \text{and} \quad \gamma_i \stackrel{\text{def}}{=} \frac{|g_i|^2}{\sigma_c^2}$$

- The values of the number of active sensors  $K_1$  for which  $a_i > 0$ , and  $\lambda_0$  are *unique* and can be found using the following equation:

$$\sum_{i=1}^{K_1} \beta_i \sqrt[3]{\frac{\beta_i \delta_i^2 T_i}{\lambda_0}} \left( 1 - \frac{2}{3} \sqrt[3]{\frac{\beta_i \delta_i^2}{\lambda_0 T_i^2}} \right) = \sum_{i=1}^{K_1} \beta_i - \frac{1}{D_0}$$

# Optimal Local Amplification Gains

The unique solution for the given constrained *convex* optimization problem could be found as follows:

$$a_i^2 = \begin{cases} \frac{1}{\gamma_i \sigma_0^2} \left( \frac{\sqrt[3]{\frac{\lambda_0}{\beta_i \delta_i^2 T_i}}}{1 - \frac{2}{3}} \sqrt[3]{\frac{\beta_i \delta_i^2}{\lambda_0 T_i^2}} - 1 \right), & i \leq K_1 \\ 0, & i > K_1 \end{cases}$$

$$T_i = 1 + \sqrt{1 + \frac{8\beta_i \delta_i^2}{27\lambda_0}} \quad \delta_i \stackrel{\text{def}}{=} \frac{1 + \beta_i}{\beta_i \gamma_i} \quad \beta_i \stackrel{\text{def}}{=} \frac{|h_i|^2}{\sigma_0^2} \quad \text{and} \quad \gamma_i \stackrel{\text{def}}{=} \frac{|g_i|^2}{\sigma_c^2}$$

- **Solution:** Water-filling-based iterative process to solve this equation:

$$\sum_{i=1}^{K_1} \beta_i \sqrt[3]{\frac{\beta_i \delta_i^2 T_i}{\lambda_0}} \left( 1 - \frac{2}{3} \sqrt[3]{\frac{\beta_i \delta_i^2}{\lambda_0 T_i^2}} \right) = \sum_{i=1}^{K_1} \beta_i - \frac{1}{D_0}$$

# Outline

- 1 Introduction and Problem Statement
- 2 System Model Description
- 3 Adaptive Optimal Power-Allocation Scheme
- 4 Limited Feedback for Adaptive Power Allocation**
- 5 Numerical Results
- 6 Conclusions



# What Does a Sensor Need to Find $a_i$ ?

$$a_i^2 = \begin{cases} \frac{1}{\gamma_i \sigma_0^2} \left( \frac{\sqrt[3]{\frac{\lambda_0}{\beta_i \delta_i^2 T_i}}}{1 - \frac{2}{3} \sqrt[3]{\frac{\beta_i \delta_i^2}{\lambda_0 T_i^2}}} - 1 \right), & i \leq K_1 \\ 0, & i > K_1 \end{cases}$$

$$\delta_i \triangleq \frac{1 + \beta_i}{\beta_i \gamma_i} \quad \beta_i \stackrel{\text{def}}{=} \frac{|h_i|^2}{\sigma_0^2} \quad \text{and} \quad \gamma_i \stackrel{\text{def}}{=} \frac{|g_i|^2}{\sigma_c^2} \quad K_1 \text{ and } \lambda_0$$

- **Assumption:** Complete *forward CSI* is available at local sensors.
  - *Not* practical in most applications, especially in large-scale WSNs.
  - Feedback information is typically transmitted through finite-rate *digital* feedback links.

# What Does a Sensor Need to Find $a_i$ ?

$$a_i^2 = \begin{cases} \frac{1}{\gamma_i \sigma_0^2} \left( \frac{\sqrt[3]{\frac{\lambda_0}{\beta_i \delta_i^2 T_i}}}{1 - \frac{2}{3}} \sqrt[3]{\frac{\beta_i \delta_i^2}{\lambda_0 T_i^2}} - 1 \right), & i \leq K_1 \\ 0, & i > K_1 \end{cases}$$

$$\delta_i \triangleq \frac{1 + \beta_i}{\beta_i \gamma_i} \quad \beta_i \stackrel{\text{def}}{=} \frac{|h_i|^2}{\sigma_0^2} \quad \text{and} \quad \gamma_i \stackrel{\text{def}}{=} \frac{|g_i|^2}{\sigma_c^2} \quad K_1 \text{ and } \lambda_0$$

- Type of required full-feedback from the FC to local sensors:
  - 1 Instantaneous amplification gain  $a_i$ .
  - 2 Fading coefficient of the channel between each sensor and the FC  $g_i$ ,  $K_1$  and  $\lambda_0$ .
    - ▶  $\lambda_0$  and an extra one-bit command instructing the sensor to transmit or stay silent.
    - ▶ Entire vector of  $\mathbf{g}$  sent by the FC over a broadcast channel.

# What Does a Sensor Need to Find $a_i$ ?

$$a_i^2 = \begin{cases} \frac{1}{\gamma_i \sigma_0^2} \left( \frac{\sqrt[3]{\frac{\lambda_0}{\beta_i \delta_i^2 T_i}}}{1 - \frac{2}{3} \sqrt[3]{\frac{\beta_i \delta_i^2}{\lambda_0 T_i}}} - 1 \right), & i \leq K_1 \\ 0, & i > K_1 \end{cases}$$

$$\delta_i \triangleq \frac{1 + \beta_i}{\beta_i \gamma_i} \quad \beta_i \stackrel{\text{def}}{=} \frac{|h_i|^2}{\sigma_0^2} \quad \text{and} \quad \gamma_i \stackrel{\text{def}}{=} \frac{|g_i|^2}{\sigma_c^2} \quad K_1 \text{ and } \lambda_0$$

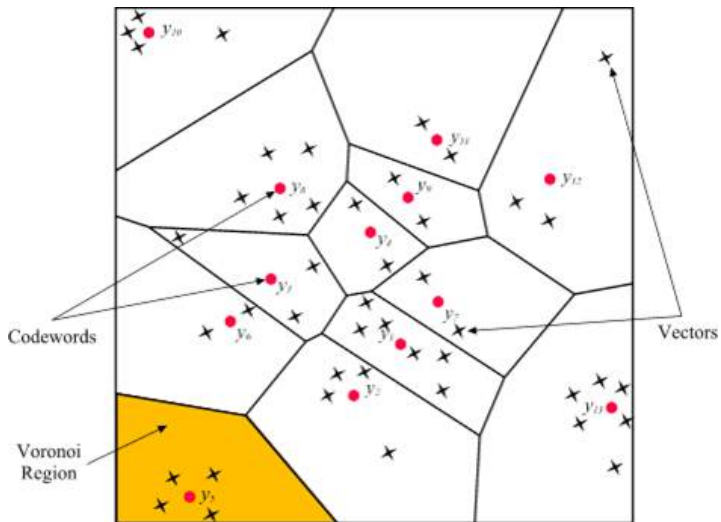
- **Solution:** Limited feedback of instantaneous amplification gain  $a_i$ 
  - For each channel realization  $\mathbf{g}$ , the FC first finds the *exact* optimal power-allocation scheme using perfect *backward* CSI.
  - The FC sends back the *index* of the quantized version of the optimized power-allocation vector to all sensors.

# Main Idea of Limited-Feedback Strategy

A *codebook* of the local amplification gains is required.

- The optimal codebook is designed *offline* by quantizing the space of the optimized power-allocation vectors using the *generalized Lloyd algorithm* with modified distortion metrics.
- $L$ : The number of *total* feedback bits from the FC to sensors.
- The space of the optimal local power-allocation vectors is quantized into  $2^L$  disjoint regions.
- A codeword is chosen in each quantization region.
- Codeword length is  $K$ . Its  $i$ th entry is a *real-valued* quantized version of the optimal local amplification gain for sensor  $i$ .

# Overview on Generalized Lloyd Quantization



# Inputs and Output of Generalized Lloyd Algorithm

- Inputs to Lloyd algorithm
  - $K$ : The number of spatially distributed sensors.
  - $L$ : The number of *total* feedback bits from the FC to sensors.
    - ▶ *Not* the number of bits fed back to each sensor.
  - Fading model of the channel between local sensors and the FC.
  - $M$ : The number of *training vectors* in space  $\mathcal{A}$ .
  - $\varepsilon$ : Distortion threshold to stop the codebook-design iterations.

# Inputs and Output of Generalized Lloyd Algorithm

- Inputs to Lloyd algorithm

- $K$ : The number of spatially distributed sensors.
- $L$ : The number of *total* feedback bits from the FC to sensors.
  - ▶ *Not* the number of bits fed back to each sensor.
- Fading model of the channel between local sensors and the FC.
- $M$ : The number of *training vectors* in space  $\mathcal{A}$ .
- $\varepsilon$ : Distortion threshold to stop the codebook-design iterations.

- Output of Lloyd algorithm

- $\mathbf{C}^{\text{OPT}}$ : Codebook matrix of the optimal local amplification gains.
- $\mathbf{C} = [\mathbf{a}_1 \ \mathbf{a}_2 \ \cdots \ \mathbf{a}_{2^L}]^T$  is a  $2^L \times K$  matrix.
- $[\mathbf{C}]_{\ell,i}$  denotes the element in row  $\ell$  and column  $i$  as the optimal gain of sensor  $i$  in codeword  $\ell$ .
- Each  $\mathbf{a}_\ell$ ,  $\ell = 1, 2, \dots, 2^L$  is associated with a realization of the fading coefficients of the channels between local sensors and the FC.

# Initialization of Generalized Lloyd Algorithm

- Generate  $\mathcal{G}_s$  as a set of  $M$  length- $K$  vectors of channel-fading realizations based on the given fading model of the channels between local sensors and the FC.
  - $M \gg 2^L$ .



# Initialization of Generalized Lloyd Algorithm

- Generate  $\mathcal{G}_s$  as a set of  $M$  length- $K$  vectors of channel-fading realizations based on the given fading model of the channels between local sensors and the FC.
  - $M \gg 2^L$ .
- Find  $\mathcal{A}_s$  as the set of optimal local power-allocation vectors associated with the channel fading vectors in  $\mathcal{G}_s$ .
  - $\mathcal{A}_s$  is the set of training vectors, and  $\mathcal{A}_s \subseteq \mathcal{A}$ .

# Initialization of Generalized Lloyd Algorithm

- Generate  $\mathcal{G}_s$  as a set of  $M$  length- $K$  vectors of channel-fading realizations based on the given fading model of the channels between local sensors and the FC.
  - $M \gg 2^L$ .
- Find  $\mathcal{A}_s$  as the set of optimal local power-allocation vectors associated with the channel fading vectors in  $\mathcal{G}_s$ .
  - $\mathcal{A}_s$  is the set of training vectors, and  $\mathcal{A}_s \subseteq \mathcal{A}$ .
- *Randomly* select  $2^L$  optimal power-allocation vectors from the set  $\mathcal{A}_s$  as the initial set of codewords.
  - Denote each codeword by  $\mathbf{a}_\ell^0$ ,  $\ell = 1, 2, \dots, 2^L$ .

# Initialization of Generalized Lloyd Algorithm

- Generate  $\mathcal{G}_s$  as a set of  $M$  length- $K$  vectors of channel-fading realizations based on the given fading model of the channels between local sensors and the FC.
  - $M \gg 2^L$ .
- Find  $\mathcal{A}_s$  as the set of optimal local power-allocation vectors associated with the channel fading vectors in  $\mathcal{G}_s$ .
  - $\mathcal{A}_s$  is the set of training vectors, and  $\mathcal{A}_s \subseteq \mathcal{A}$ .
- *Randomly* select  $2^L$  optimal power-allocation vectors from the set  $\mathcal{A}_s$  as the initial set of codewords.
  - Denote each codeword by  $\mathbf{a}_\ell^0$ ,  $\ell = 1, 2, \dots, 2^L$ .
- Set the initial codebook as  $\mathbf{C}^0 \leftarrow [\mathbf{a}_1^0 \ \mathbf{a}_2^0 \ \cdots \ \mathbf{a}_{2^L}^0]^T$ .
- Find the *average distortion of codebook*.
  - Set  $\text{NewCost} \leftarrow D_B(\mathbf{C}^0)$  and  $j \leftarrow 0$ .

# Modified Distortions for Generalized Lloyd Algorithm

- Let  $\mathbf{P}_\ell$  and  $\mathbf{P}$  be the vectors of local transmit powers, when the vector of local amplification gains is  $\mathbf{a}_\ell$  and  $\mathbf{a}$ , respectively.

$$P_i = a_i^2 \left( |h_i|^2 + \sigma_o^2 \right) = a_i^2 \sigma_o^2 (1 + \beta_i)$$

# Modified Distortions for Generalized Lloyd Algorithm

- Let  $\mathbf{P}_\ell$  and  $\mathbf{P}$  be the vectors of local transmit powers, when the vector of local amplification gains is  $\mathbf{a}_\ell$  and  $\mathbf{a}$ , respectively.

$$P_i = a_i^2 \left( |h_i|^2 + \sigma_o^2 \right) = a_i^2 \sigma_o^2 (1 + \beta_i)$$

- Optimization cost of a power-allocation vector

$$J(\mathbf{a}) = \left( \sum_{i=1}^K P_i^2 \right)^{\frac{1}{2}} = \left( \sum_{i=1}^K [a_i^2 \sigma_o^2 (1 + \beta_i)]^2 \right)^{\frac{1}{2}}$$

# Modified Distortions for Generalized Lloyd Algorithm

- Let  $\mathbf{P}_\ell$  and  $\mathbf{P}$  be the vectors of local transmit powers, when the vector of local amplification gains is  $\mathbf{a}_\ell$  and  $\mathbf{a}$ , respectively.
- Optimization cost of a power-allocation vector

$$J(\mathbf{a}) = \left( \sum_{i=1}^K P_i^2 \right)^{\frac{1}{2}} = \left( \sum_{i=1}^K [a_i^2 \sigma_o^2 (1 + \beta_i)]^2 \right)^{\frac{1}{2}}$$

- Distance between codeword  $\mathbf{a}_\ell$  and the optimal power-allocation vector  $\mathbf{a}$

$$D_W(\mathbf{a}_\ell, \mathbf{a}) \triangleq |J(\mathbf{a}_\ell) - J(\mathbf{a})|$$

# Modified Distortions for Generalized Lloyd Algorithm

- Let  $\mathbf{P}_\ell$  and  $\mathbf{P}$  be the vectors of local transmit powers, when the vector of local amplification gains is  $\mathbf{a}_\ell$  and  $\mathbf{a}$ , respectively.
- Optimization cost of a power-allocation vector

$$J(\mathbf{a}) = \left( \sum_{i=1}^K P_i^2 \right)^{\frac{1}{2}} = \left( \sum_{i=1}^K [a_i^2 \sigma_o^2 (1 + \beta_i)]^2 \right)^{\frac{1}{2}}$$

- Distance between codeword  $\mathbf{a}_\ell$  and the optimal power-allocation vector  $\mathbf{a}$

$$D_W(\mathbf{a}_\ell, \mathbf{a}) \triangleq |J(\mathbf{a}_\ell) - J(\mathbf{a})|$$

- Average distortion for codebook  $\mathbf{C}$

$$D_B(\mathbf{C}) \triangleq \mathbb{E}_{\mathbf{a}} \left[ \min_{\ell \in \{1, 2, \dots, 2^L\}} D_W(\mathbf{a}_\ell, \mathbf{a}) \right]$$

# Detailed Operation of Generalized Lloyd Algorithm

Repeat the following steps:

- $j \leftarrow j + 1$  and  $\text{OldCost} \leftarrow \text{NewCost}$ .



# Detailed Opertaion of Generalized Lloyd Algorithm

Repeat the following steps:

- $j \leftarrow j + 1$  and OldCost  $\leftarrow$  NewCost.
- Given codebook  $\mathbf{C}^{j-1}$ , optimally partition the set  $\mathcal{A}_s$  into  $2^L$  disjoint subsets and denote the resulted optimal partitions by  $\mathcal{A}_\ell^{j-1}$ ,  $\ell = 1, 2, \dots, 2^L$ .
  - **Nearest-Neighbor Condition** finds the optimal *Voronoi cells* of the vector space to be quantized, given a fixed codebook.
  - Each point  $\mathbf{a} \in \mathcal{A}_s$  is assigned to partition  $\ell$  represented by codeword  $\mathbf{a}_\ell \in \mathbf{C}^{j-1}$  if and only if its distance to codeword  $\mathbf{a}_\ell$ , with respect to the defined distance function, is less than its distance to any other codeword in the codebook.
  - Given a codebook  $\mathbf{C}^{j-1}$ , the space  $\mathcal{A}_s$  is divided into  $2^L$  disjoint quantization regions (or Voronoi cells) with the  $\ell$ th region defined as

$$\mathcal{A}_\ell = \{\mathbf{a} \in \mathcal{A}_s : D_W(\mathbf{a}_\ell, \mathbf{a}) \leq D_W(\mathbf{a}_k, \mathbf{a}), \forall k \neq \ell\}$$

# Detailed Opertaion of Generalized Lloyd Algorithm

Repeat the following steps:

- $j \leftarrow j + 1$  and OldCost  $\leftarrow$  NewCost.
- Given codebook  $\mathbf{C}^{j-1}$ , optimally partition the set  $\mathcal{A}_s$  into  $2^L$  disjoint subsets and denote the resulted optimal partitions by  $\mathcal{A}_\ell^{j-1}$ .
- For every  $\mathcal{A}_\ell^{j-1}$ ,  $\ell = 1, 2, \dots, 2^L$ :
  - Find the optimal codeword associated with partition  $\mathcal{A}_\ell^{j-1}$ . Denote this new optimal codeword as  $\mathbf{a}_\ell^j$ .

# Detailed Opertaion of Generalized Lloyd Algorithm

Repeat the following steps:

- $j \leftarrow j + 1$  and OldCost  $\leftarrow$  NewCost.
- Given codebook  $\mathbf{C}^{j-1}$ , optimally partition the set  $\mathcal{A}_s$  into  $2^L$  disjoint subsets and denote the resulted optimal partitions by  $\mathcal{A}_\ell^{j-1}$ .
- Find the optimal codeword associated with partition  $\mathcal{A}_\ell^{j-1}$ .
  - **Centroid Condition** finds the optimal codebook, given a specific partitioning of the vector space.
  - The optimal codeword associated with each Voronoi cell  $\mathcal{A}_\ell \subseteq \mathcal{A}_s$  is the *centroid* of that cell with respect to the defined distance function.
  - Given a specific partitioning  $\{\mathcal{A}_1, \mathcal{A}_2, \dots, \mathcal{A}_{2^L}\}$ , the optimal codeword associated with partition  $\mathcal{A}_\ell$  is defined as

$$\mathbf{a}_\ell^* = \arg \min_{\mathbf{a}_\ell \in \mathcal{A}_\ell} \mathbb{E}_{\mathbf{a} \in \mathcal{A}_\ell} [D_W(\mathbf{a}_\ell, \mathbf{a})]$$

# Detailed Opertaion of Generalized Lloyd Algorithm

Repeat the following steps:

- $j \leftarrow j + 1$  and OldCost  $\leftarrow$  NewCost.
- Given codebook  $\mathbf{C}^{j-1}$ , optimally partition the set  $\mathcal{A}_s$  into  $2^L$  disjoint subsets and denote the resulted optimal partitions by  $\mathcal{A}_\ell^{j-1}$ .
- For every  $\mathcal{A}_\ell^{j-1}$ ,  $\ell = 1, 2, \dots, 2^L$ :
  - Find the optimal codeword associated with partition  $\mathcal{A}_\ell^{j-1}$ . Denote this new optimal codeword as  $\mathbf{a}_\ell^j$ .
- Form the new codebook  $\mathbf{C}^j \leftarrow \left[ \mathbf{a}_1^j \ \mathbf{a}_2^j \ \dots \ \mathbf{a}_{2^L}^j \right]^T$ .
  - Set NewCost  $\leftarrow D_B(\mathbf{C}^j)$ .

# Detailed Opertaion of Generalized Lloyd Algorithm

Repeat the following steps:

- $j \leftarrow j + 1$  and OldCost  $\leftarrow$  NewCost.
- Given codebook  $\mathbf{C}^{j-1}$ , optimally partition the set  $\mathcal{A}_s$  into  $2^L$  disjoint subsets and denote the resulted optimal partitions by  $\mathcal{A}_\ell^{j-1}$ .
- For every  $\mathcal{A}_\ell^{j-1}$ ,  $\ell = 1, 2, \dots, 2^L$ :
  - Find the optimal codeword associated with partition  $\mathcal{A}_\ell^{j-1}$ . Denote this new optimal codeword as  $\mathbf{a}_\ell^j$ .
- Form the new codebook  $\mathbf{C}^j \leftarrow \left[ \mathbf{a}_1^j \ \mathbf{a}_2^j \ \dots \ \mathbf{a}_{2^L}^j \right]^T$ .
  - Set NewCost  $\leftarrow D_B(\mathbf{C}^j)$ .

Stop if OldCost – NewCost  $\leq \varepsilon$ .

# Comments on Generalized Lloyd Algorithm

- The average codebook distortion monotonically decreases through the iterative usage of the Centroid Condition and the Nearest Neighbor Condition.
- The optimal codebook is designed offline by the FC once.
- The optimal codebook is stored in the FC and all sensors.

# Comments on Generalized Lloyd Algorithm

- Upon observing a realization of the channel vector  $\mathbf{g}$ :
  - The FC finds the associated optimal power-allocation vector  $\mathbf{a}^{\text{OPT}}$ .
  - It identifies the closest codeword in the optimal codebook  $\mathbf{C}$  to  $\mathbf{a}^{\text{OPT}}$  with respect to the defined distance metric.

$$\ell = \arg \min_{k \in \{1, 2, \dots, 2^L\}, \mathbf{a}_k \in \mathbf{C}} D_{\text{W}}(\mathbf{a}_k, \mathbf{a}^{\text{OPT}}).$$

- It broadcasts the *index* of that codeword  $\ell$  to all sensors.

# Comments on Generalized Lloyd Algorithm

- Upon observing a realization of the channel vector  $\mathbf{g}$ :
  - The FC finds the associated optimal power-allocation vector  $\mathbf{a}^{\text{OPT}}$ .
  - It identifies the closest codeword in the optimal codebook  $\mathbf{C}$  to  $\mathbf{a}^{\text{OPT}}$  with respect to the defined distance metric.

$$\ell = \arg \min_{k \in \{1, 2, \dots, 2^L\}, \mathbf{a}_k \in \mathbf{C}} D_{\mathbf{W}}(\mathbf{a}_k, \mathbf{a}^{\text{OPT}}).$$

- It broadcasts the *index* of that codeword  $\ell$  to all sensors.
- Upon reception of the index  $\ell$ , the quantized local amplification gain for each sensor  $i$  is  $[\mathbf{C}]_{\ell, i}$ .



# Outline

- 1 Introduction and Problem Statement
- 2 System Model Description
- 3 Adaptive Optimal Power-Allocation Scheme
- 4 Limited Feedback for Adaptive Power Allocation
- 5 Numerical Results**
- 6 Conclusions

# Simulation Setup

- Local observation gains are randomly chosen from a Gaussian distribution  $h_i \sim \mathcal{N}(1, 0.09)$ .
- Average power of  $h_i$  across all sensors is 1.2.
- Observation noise variance is  $\sigma_o^2 = 10$  dBm.
- Channel noise variance is  $\sigma_c^2 = -90$  dBm.
- Fading model for the channels between local sensors and the FC:

$$g_i = \eta_0 \left( \frac{d_i}{d_0} \right)^{-\alpha} f_i$$

- $\eta_0 = -30$  dB: Nominal fading gain at reference distance  $d_0 = 1$  meter.
- $d_i$ : Distance between sensor  $i$  and the FC (in meters).
- $\alpha = 2$ : Path-loss exponent.
- $f_i$ : i.i.d. Rayleigh-fading random variable with unit variance.
- The distance between sensors and the FC is uniformly distributed between 50 and 150 meters.

# Simulation Setup

- Local observation gains are randomly chosen from a Gaussian distribution  $h_i \sim \mathcal{N}(1, 0.09)$ .
- Average power of  $h_i$  across all sensors is 1.2.
- Observation noise variance is  $\sigma_o^2 = 10$  dBm.
- Channel noise variance is  $\sigma_c^2 = -90$  dBm.
- Fading model for the channels between local sensors and the FC:

$$g_i = \eta_0 \left( \frac{d_i}{d_0} \right)^{-\alpha} f_i, \quad i = 1, 2, \dots, K,$$

- The training-set size in the optimal codebook-design is  $M = 5,000$ .
- Codebook-distortion threshold is  $\varepsilon = 10^{-4}$ .
- The results are obtained by averaging over 10,000 Monte Carlo simulations.

# Simulation Setup

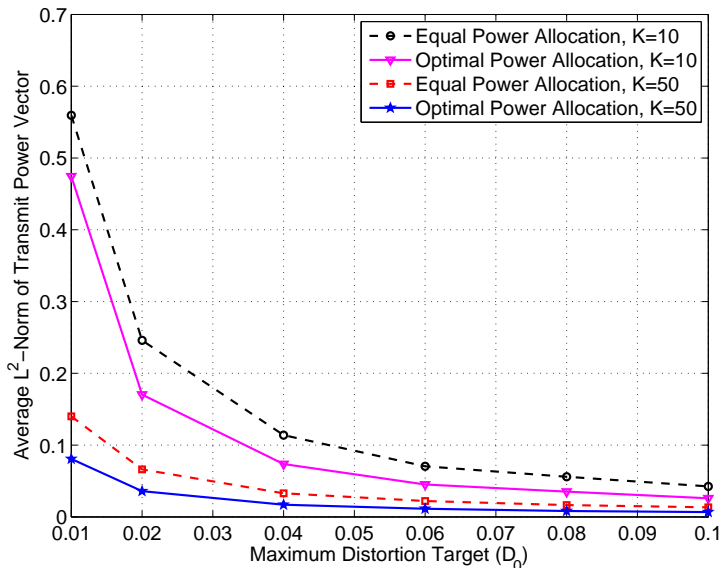
- Local observation gains are randomly chosen from a Gaussian distribution  $h_i \sim \mathcal{N}(1, 0.09)$ .
- Average power of  $h_i$  across all sensors is 1.2.
- Observation noise variance is  $\sigma_o^2 = 10$  dBm.
- Channel noise variance is  $\sigma_c^2 = -90$  dBm.
- Fading model for the channels between local sensors and the FC:

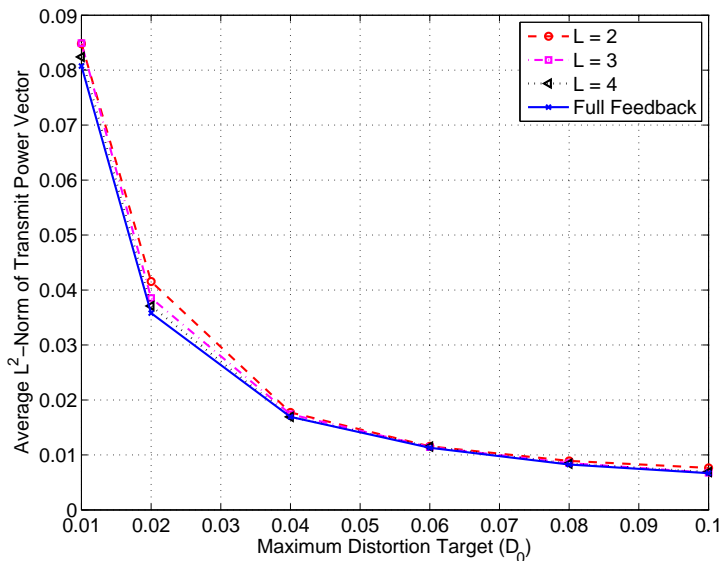
$$g_i = \eta_0 \left( \frac{d_i}{d_0} \right)^{-\alpha} f_i, \quad i = 1, 2, \dots, K,$$

- Performance Metric:** The energy efficiency of a power-allocation scheme defined as

$$J(\mathbf{a}) = \left( \sum_{i=1}^K P_i^2 \right)^{\frac{1}{2}} = \left( \sum_{i=1}^K [a_i^2 \sigma_o^2 (1 + \beta_i)]^2 \right)^{\frac{1}{2}}$$

## Energy Efficiency of Adaptive Power Allocation



Effect of  $L$  on Energy Efficiency ( $K = 50$  Sensors)

# Outline

- 1 Introduction and Problem Statement
- 2 System Model Description
- 3 Adaptive Optimal Power-Allocation Scheme
- 4 Limited Feedback for Adaptive Power Allocation
- 5 Numerical Results
- 6 Conclusions**

# Summary

- An adaptive power-allocation scheme was proposed for distributed BLUE estimation of a random scalar parameter at the FC of a WSN.
- The proposed scheme minimizes the  $L^2$ -norm of the vector of local transmit powers, given a maximum estimation distortion measured as the variance of the BLUE estimator.
- A limited-feedback strategy was proposed to eliminate the requirement of infinite-rate feedback of the instantaneous forward CSI from the FC to local sensors.
  - Designed an optimal codebook by quantizing the vector space of the optimal local amplification gains using the generalized Lloyd algorithm with modified distortion functions.



**Thank You for Your Attention.**

Questions and/or Comments?