

Coherent and Multi-symbol Noncoherent CPFSK: Capacity and Code Design

Shi Cheng¹ Matthew C. Valenti¹ Don Torrieri²

¹Lane Department of Computer Science and Electrical Engineering
West Virginia University

²US Army Research Lab

October 31, 2007

Outline

1 Coherent CPFSK

- System Model and Trellis Decoding
- Capacity of Coherent CPFSK
- Capacity Approaching Code Design

2 Multi-symbol Noncoherent CPFSK

- Capacity of Multi-symbol Noncoherent CPFSK
- Code Design

3 Conclusion

Discrete Time Model

$$\mathbf{y} = e^{j\phi} \sqrt{\mathcal{E}_s} \mathbf{x} + \mathbf{n}$$

- \mathbf{y} is the output from M complex filters matched to the tones,
- \mathbf{x} is chosen from columns of $\mathbf{K} = [\mathbf{k}_0, \mathbf{k}_1, \dots, \mathbf{k}_{M-1}]$
- \mathbf{n} is colored noise, with $E(\mathbf{n}\mathbf{n}^H) = N_0\mathbf{K}$.
- Decoding metric

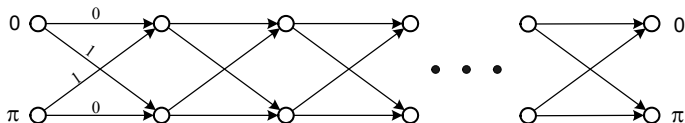
$$\log p(\mathbf{y}|\mathbf{x} = \mathbf{k}_\nu, \phi) = 2 \frac{\sqrt{\mathcal{E}_s}}{N_0} \text{Re}(e^{-j\phi} y_\nu) + \text{Constant}.$$

CPFSK Trellis

- $h = \frac{P}{Q}$, which means the trellis has Q states
- Suppose the initial phase is 0,

$$\phi_i \in \Phi = \left\{ \frac{2k\pi}{Q}, k = 0, 1, \dots, Q-1 \right\}$$

- Example of MSK: $M = 2$, $h = 1/2$



Capacity of Coherent CPFSK Detection

- i.u.d. capacity calculation through trellis

$$\begin{aligned}
 C^{(c)} &= \lim_{N \rightarrow \infty} \frac{1}{N} I(\mathbf{x}_0^{N-1}, \mathbf{y}_0^{N-1}) \\
 &= \lim_{N \rightarrow \infty} \frac{1}{N} \left[\sum_{i=0}^{N-1} H(\mathbf{x}_i) - \sum_{i=0}^{N-1} H(\mathbf{x}_i | \mathbf{x}_0^{i-1}, \mathbf{y}_0^{N-1}) \right]
 \end{aligned}$$

- Treating CPFSK as a finite state markov channel (FSMC)
- Using Monte Carlo simulation to find $E \left[-\log p(\mathbf{x}_i | \mathbf{x}_0^{i-1}, \mathbf{y}_0^{N-1}) \right]$

Capacity of Coherent CPFSK Detection

- i.u.d. capacity calculation through trellis

$$\begin{aligned}
 C^{(c)} &= \lim_{N \rightarrow \infty} \frac{1}{N} I(\mathbf{x}_0^{N-1}, \mathbf{y}_0^{N-1}) \\
 &= \lim_{N \rightarrow \infty} \frac{1}{N} \left[\sum_{i=0}^{N-1} H(\mathbf{x}_i) - \sum_{i=0}^{N-1} H(\mathbf{x}_i | \mathbf{x}_0^{i-1}, \mathbf{y}_0^{N-1}) \right]
 \end{aligned}$$

- Treating CPFSK as a finite state markov channel (FSMC)
- Using Monte Carlo simulation to find $E \left[-\log p(\mathbf{x}_i | \mathbf{x}_0^{i-1}, \mathbf{y}_0^{N-1}) \right]$

Capacity of Coherent CPFSK Detection

- i.u.d. capacity calculation through trellis

$$\begin{aligned}
 C^{(c)} &= \lim_{N \rightarrow \infty} \frac{1}{N} I(\mathbf{x}_0^{N-1}, \mathbf{y}_0^{N-1}) \\
 &= \lim_{N \rightarrow \infty} \frac{1}{N} \left[\sum_{i=0}^{N-1} H(\mathbf{x}_i) - \sum_{i=0}^{N-1} H(\mathbf{x}_i | \mathbf{x}_0^{i-1}, \mathbf{y}_0^{N-1}) \right]
 \end{aligned}$$

- Treating CPFSK as a finite state markov channel (FSMC)
- Using Monte Carlo simulation to find $E \left[-\log p(\mathbf{x}_i | \mathbf{x}_0^{i-1}, \mathbf{y}_0^{N-1}) \right]$

BCJR Algorithm

- Finding $p(\mathbf{x}_i | \mathbf{y}_0^{N-1})$
- Define α , β and γ as

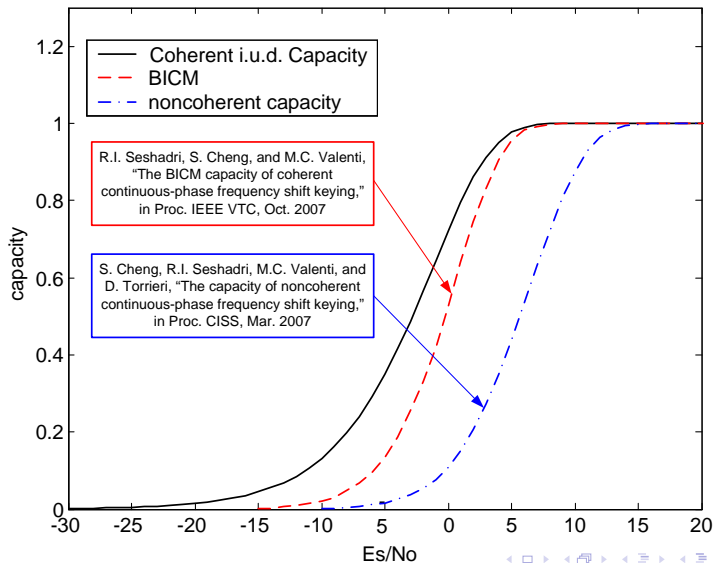
$$\begin{aligned} \alpha_i(\phi_i) &\triangleq p(\phi_i, \mathbf{y}_0^{i-1}) \\ \beta_{i+1}(\phi_{i+1}) &\triangleq p(\mathbf{y}_{i+1}^{N-1} | \phi_{i+1}) \\ \gamma(\phi_i \rightarrow \phi_{i+1}, \mathbf{y}_i) &\triangleq p(\mathbf{y}_i, \phi_{i+1} | \phi_i). \end{aligned}$$

Modification to BCJR Algorithm

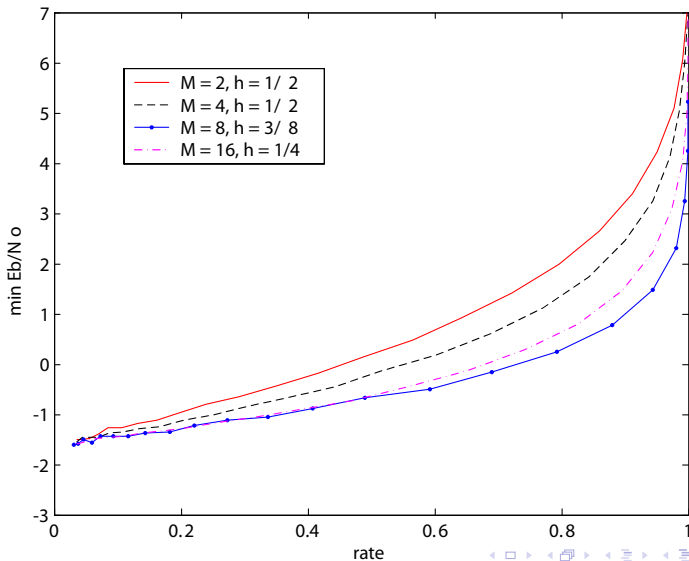
- Finding $p(\mathbf{x}_i | \mathbf{x}_0^{i-1}, \mathbf{y}_0^{N-1})$
- Define α , β and γ as

$$\begin{aligned} \alpha_i(\phi_i) &\triangleq p(\phi_i, \mathbf{y}_0^{i-1}, \mathbf{x}_0^{i-1}) \\ \beta_{i+1}(\phi_{i+1}) &\triangleq p(\mathbf{y}_{i+1}^{N-1} | \phi_{i+1}) \\ \gamma(\phi_i \rightarrow \phi_{i+1}, \mathbf{y}_i, \mathbf{x}_i) &\triangleq p(\mathbf{y}_i, \phi_{i+1} | \phi_i, \mathbf{x}_i). \end{aligned}$$

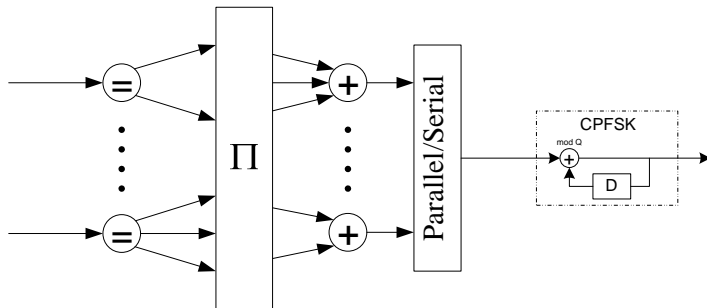
Capacity Curves: MSK



Coherent Capacity Curves

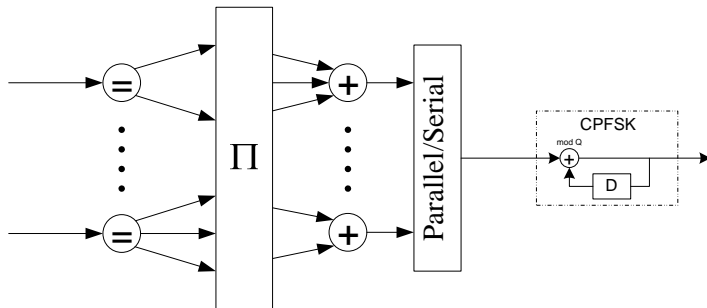


Code Design



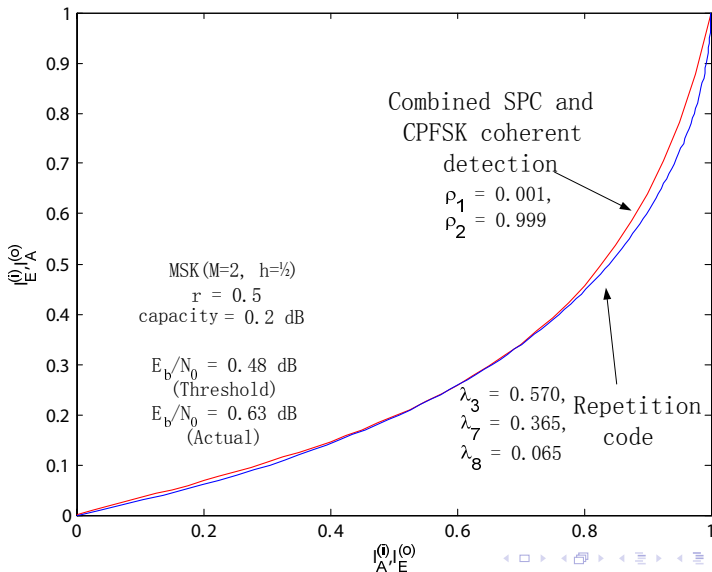
- Irregular repeat-accumulate (IRA) code structure
- Using linear programming to design capacity approaching code

Code Design



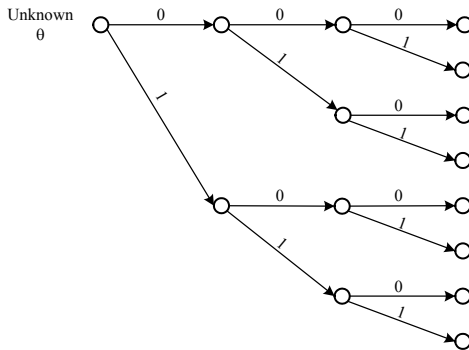
- Irregular repeat-accumulate (IRA) code structure
- Using linear programming to design capacity approaching code

EXIT Curves

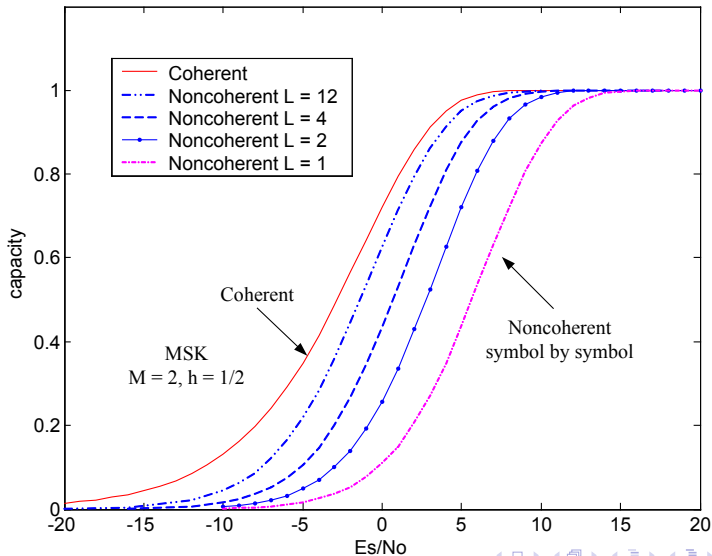


Multi-symbol Noncoherent Detector

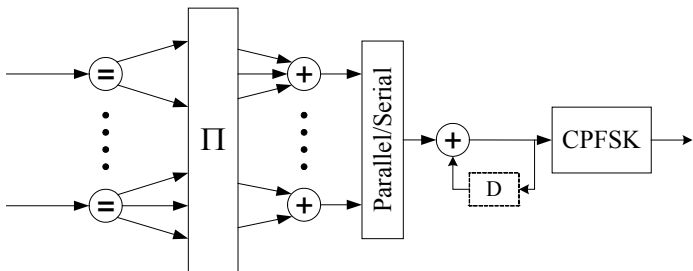
- Noncoherent detector over L symbols
- Approach capacity of coherent detector when L is large enough
- Detector structure



Multi-symbol Noncoherent Detector vs Coherent (MSK)

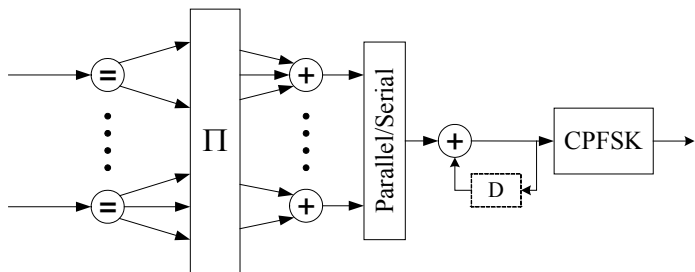


IRA Coded System



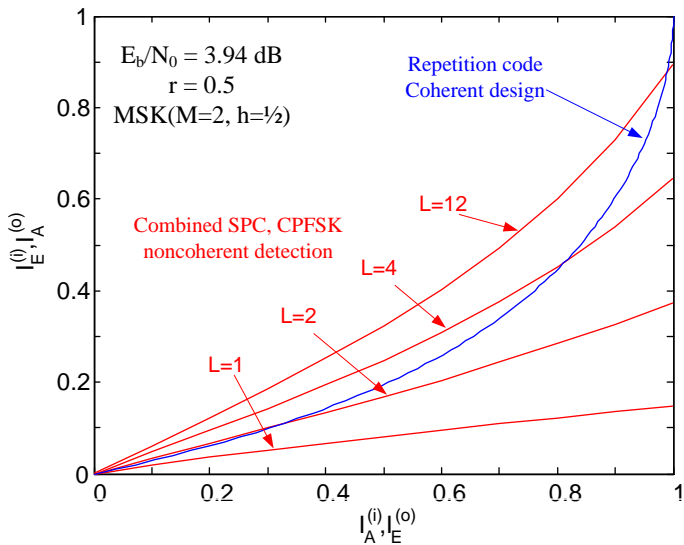
- Irregular repeat-accumulate (IRA) code structure
- Using differential encoder

IRA Coded System

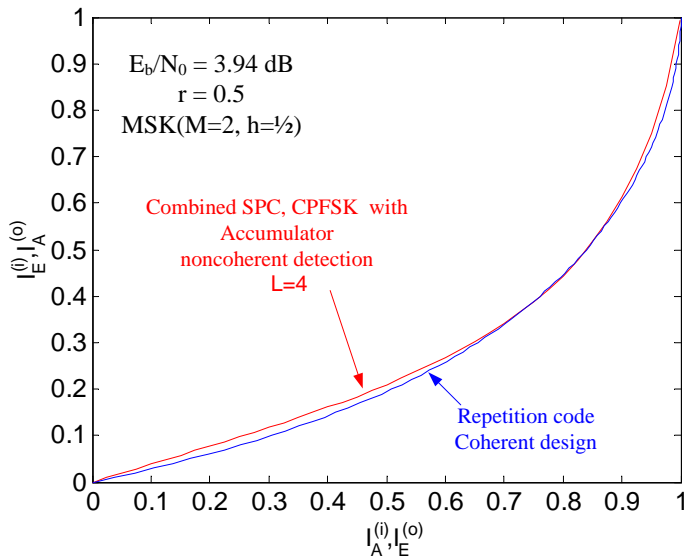


- Irregular repeat-accumulate (IRA) code structure
- Using differential encoder

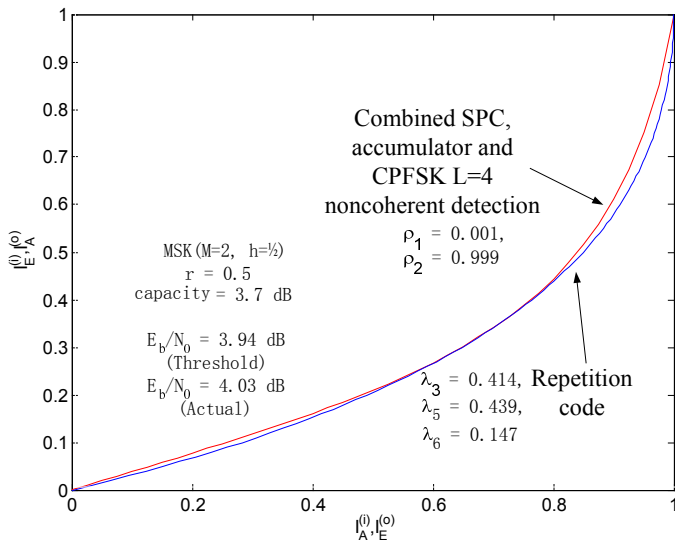
Without Accumulator



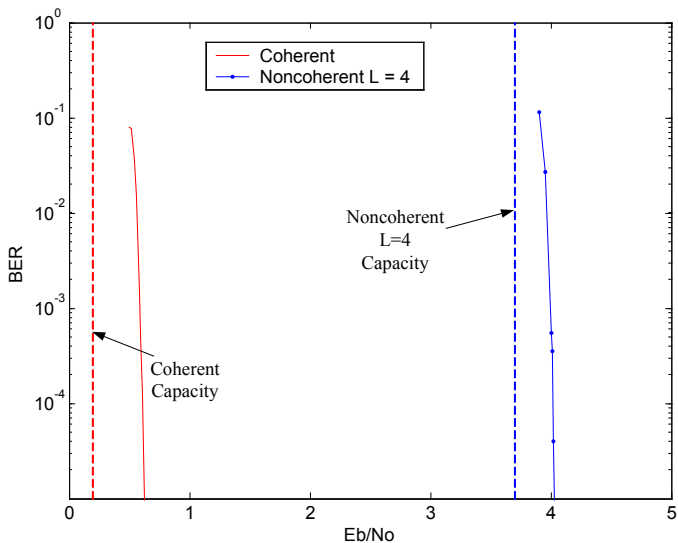
With Accumulator



A New Design



BER Curves



Conclusion

- The i.u.d. capacity of coherent detected CPFSK in AWGN channel is found by treating CPFSK as a finite-state Markov channel.
- The capacity of multi-symbol noncoherent detected CPFSK is studied. When the block size L increases, this capacity gets closer to the coherent capacity. For example, at coding rate $r = 0.5$, the $L = 4$ noncoherent MSK detector is $3.5dB$ worse than the coherent one, but $5dB$ better than the symbol-by-symbol noncoherent detector.
- Capacity approaching codes are designed for both detectors. IRA code structure is utilized, and the degree distribution is optimized through the linear programming EXIT curve fitting approach.

Thank you