# Adjacent-Channel Interference in Frequency-Hopping Ad Hoc Networks

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- 1 Frequency-Hopping Ad Hoc Networks
- Outage Probability with Adjacent Channel Interference
- 3 Modulation-Constrained Transmission Capacity
- Optimization Results
- 5 Conclusions

### Outline

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## Ad Hoc Networks



- Mobile transmitters are randomly placed in a 2-D finite space.
- A reference receiver is located at the origin.
- $X_i$  represents the  $i^{th}$  transmitter and its location.
  - $X_0$  is location of the reference transmitter.
  - M interfering transmitters,  $\{X_1, ..., X_M\}$ .
  - $|X_i|$  is distance from  $i^{th}$  transmitter to the reference receiver.
- Spatial model
  - Fixed number of interferers, placed independently and uniformly over the network area (BPP).

## Frequency Hopping

• To manage interference, *frequency hopping* (FH) is used:



- Each mobile transmits with *duty factor* D.
  - Likelihood of a transmission is  $D \leq 1$ .
- $\bullet\,$  Each transmitting mobile randomly picks from among L frequencies.
  - Probability of a *co-channel collision* is  $p_c = D/L$ .

# Adjacent-Channel Interference

- We assumed that all interference is co-channel interference.
  - However, FSK modulation is not completely bandlimited.
  - Some of the spectrum splatters into adjacent channels.
- Typically, the frequency channels are matched to the 99-percent bandwidth of the modulation.
  - Percent of the signal power splatters into adjacent channels.
  - Containing 99-percent of the signal power in the channel is arbitrary.
- There is a tradeoff in the choice of percent power.
  - Let  $\psi < 1$  be the fraction of power in the band.
  - Then the fraction of power in each adjacent channel is  $K_s = (1 \psi)/2$ .
- Each transmitting mobile has probability of an *adjacent-channel collision* given by

$$p_{\mathsf{a}} = D\left[\left(\frac{2}{L}\right)\left(\frac{L-2}{L}\right) + \left(\frac{1}{L}\right)\left(\frac{2}{L}\right)\right] = \frac{2D(L-1)}{L^2}.$$

## SINR

The performance at the reference receiver is characterized by the *signal-to-interference and noise ratio* (SINR), given by:

$$\gamma = \frac{g_0 \Omega_0}{\Gamma^{-1} + \sum_{i=1}^M I_i g_i \Omega_i}$$
(1)

where:

•  $\Gamma$  is the SNR at unit distance.

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- g<sub>i</sub> is the power gain due to Nakagami fading.
- $I_i$  is a discrete random variable that characterizes the type of collision:

$$I_{i} = \begin{cases} \psi & \text{with probability } p_{c} \text{ for co-channel collision} \\ K_{s} & \text{with probability } p_{a} \text{ for adjacent-channel collision} \\ 0 & \text{with probability } p_{n} = 1 - p_{c} - p_{a} \text{ for no collision} \end{cases}$$
(2)

•  $\Omega_i = \frac{P_i}{P_0} 10^{\xi_i/10} ||X_i||^{-\alpha}$  is the normalized receiver power from transmitter i in presence of log-normal shadowing.

- $P_i$  is the power of transmitter *i*, assumed to be constant for  $\forall i$ .
- $\alpha$  is the path loss.
- $||X_0|| = 1$  to normalize distance.
- $\xi_i$  are i.i.d. zero-mean Gaussian with standard deviation  $\sigma_s$  dB.

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## **Outage Probability**

- An *outage* occurs when the SINR is below a threshold  $\beta$ .
  - $\beta$  depends on the choice of modulation and coding.
- The outage probability is

$$\epsilon = P[\gamma \le \beta]. \tag{3}$$

• Substituting (1) into (3) and rearranging yields

$$\epsilon = P \bigg[ \underbrace{\beta^{-1} g_0 \Omega_0 - \sum_{i=1}^M I_i g_i \Omega_i}_{\mathsf{Z}} \le \Gamma^{-1} \bigg].$$

• The outage probability is related to the cdf of Z,

$$\epsilon = P\left[\mathsf{Z} \le \Gamma^{-1}\right] = F_{\mathsf{Z}}(\Gamma^{-1}).$$

## **Outage Probability Derivation**

The cdf conditioned on the network geometry,  $\boldsymbol{\Omega}$ :  $\epsilon_o = P\left[\gamma \leq \beta | \boldsymbol{\Omega} \right] = F_{\mathsf{Z}}(\Gamma^{-1} | \boldsymbol{\Omega})$ 

## **Outage Probability Derivation**



#### Conditional Outage Probability

The outage probability conditioned on the network geometry:

$$\bar{F}_{\mathsf{Z}}(z|\mathbf{\Omega}) = e^{-\beta_0 z} \sum_{j=0}^{m_0-1} (\beta_0 z)^j \sum_{k=0}^j \frac{z^{-k} H_k(\mathbf{\Omega})}{(j-k)!}$$
(5)

where  $\beta_0 = m_0 \beta / (\Psi \Omega_0)$ ,

$$H_k(\mathbf{\Omega}) = \sum_{\substack{\ell_i \ge 0\\\sum_{i=0}^M \ell_i = k}} \prod_{i=1}^M G_{\ell_i}(\Omega_i),$$
(6)

the summation in (6) is over all sets of positive indices that sum to k, and

$$G_{\ell_i}(\Omega_i) = p_{\mathsf{n}} \delta_{\ell_i} + \frac{\Gamma(\ell_i + m_i)}{\ell_i! \Gamma(m_i)} \left[ p_{\mathsf{c}} \phi_i(\psi) + p_{\mathsf{a}} \phi_i(K_s) \right],$$

 $\delta_\ell$  is the Kronecker delta function, and

$$\phi_i(x) = \left(\frac{x\Omega_i}{m_i}\right)^{\ell_i} \left(\frac{x\beta_0\Omega_i}{m_i} + 1\right)^{-(m_i + \ell_i)}$$

## **Conditional Outage Probability**



Figure: Conditional outage probability as a function of SNR  $\Gamma$ . Example:

- M = 50 interferers.
- Annular network ( $r_{ex} = 0.25$  and  $r_{net} = 4$ ).
- α = 3.
- L = 200 and D = 1.
- $\beta = 3.7 \text{ dB}.$
- Analytical curves are the lines, while represents simulated values.

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## Average Outage Probability



Figure: Average outage probability as a function of the number of interferers M for two values of L. For each L, curves are shown for the case of no shadowing, and for shadowing with three values of  $\sigma_s$ .

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## Transmission Capacity: Definition

- The transmission capacity (TC) is the spatial spectral efficiency.
- If there are  $\lambda$  mobiles per unit area, then the number of successful transmissions per unit area is

$$\tau = \lambda(1-\epsilon)$$

• If the outage probability  $\epsilon$  is constrained to not exceed  $\zeta,$  then the transmission capacity is

$$\tau_c(\zeta) = \epsilon^{-1}(\zeta)(1-\zeta) \tag{7}$$

where  $\epsilon^{-1}(\zeta)$  is the maximum mobile density such that  $\epsilon \leq \zeta$ .

## Modulation-Constrained TC

• The modulation-constrained transmission capacity (MCTC) is

$$\tau' \quad = \quad \lambda(1-\epsilon) \left( \frac{RD\eta(h,\Psi)}{L} \right)$$

where

- $\bullet \ R$  is the rate of the channel code.
- *h* is the modulation index.
- $\eta(h,\Psi)$  the modulation's spectral efficiency (bps/Hz).
- $\epsilon = P[C(\gamma) \leq R] = P[\gamma \leq C^{-1}(R)]$ , where  $C(\gamma)$  is the modulation-constrained capacity with SNR  $\gamma$ .
- L is the number of channels.
- D is the duty factor.
- $\lambda$  is the density of mobiles.
- $\tau'$  has units of  $bps/Hz/m^2$ .
- For a given  $\lambda,$   $\Gamma,$  and spatial model, there is a set of  $(L,R,h,\Psi)$  that maximizes  $\tau'.$

## Noncoherent Binary CPFSK

FH systems often use noncoherent continuous-phase frequency-shift keying (CPFSK).

- The modulation-constrained capacity  $C(h, \gamma)$  of binary CPFSK is shown as a function of  $\gamma$  in figure (a).
- The bandwidth  $B = 1/\eta(h, \Psi)$  is shown as function of the modulation index h in figure (b).



(b) bandwidth versus modulation index

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## **Optimization Objectives**

- For a given network model  $(r_{ex}, r_{net}, M)$  and channel, that takes in account for shadowing  $(\sigma_s)$ , path loss  $(\alpha)$ , Nakagami fading  $(m_0 \text{ and } m_i)$  and channel SNR  $\Gamma$ , the values of  $(L, R, h, \Psi)$  that maximize the modulation-constrained transmission capacity  $\tau'$  are found.
- Optimization is through a simplex-search strategy (Nelder-Mead method) since the optimization surface is convex.



### Example of optimization

$r_{net}$	$\sigma_{\rm s}$	Fading	L	R	h	$\psi$	$ au_{opt}'$
2	0	Rayleigh	36	0.64	0.81	0.96	17.74
		Nakagami	45	0.64	0.81	0.96	19.88
		Mixed	40	0.64	0.81	0.96	22.59
	8	Rayleigh	34	0.64	0.81	0.96	18.29
		Nakagami	44	0.66	0.81	0.96	19.36
		Mixed	38	0.64	0.81	0.96	22.09
4	0	Rayleigh	13	0.57	0.85	0.95	11.92
		Nakagami	16	0.54	0.85	0.95	13.23
		Mixed	15	0.56	0.85	0.95	14.64
	8	Rayleigh	12	0.58	0.85	0.95	12.22
		Nakagami	15	0.54	0.85	0.95	13.13
		Mixed	14	0.57	0.85	0.95	14.55

Table: Results of the optimization for M = 50 interferers and  $\Gamma = 10$  dB. The normalized MCTC  $\tau'$  is in units of bps/kHz- $m^2$ .

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- The performance of frequency-hopping ad hoc networks is a function of:
  - Number of hopping channels *L*.
  - Code rate R.
  - Modulation index *h* (if CPFSK modulation).
  - Fractional in-band power  $\Psi$ .
- These parameters can be jointly optimized.
  - Transmission capacity is the objective function of choice.
  - The modulation-constrained TC quantifies the tradeoffs involved.
- The approach is general enough to handle a wide variety of conditions. Examples of future extensions:
  - Any spatial model, including repulsion models.
  - Adaptive code rates (*R* not fixed for all users).
- Additional constraints can be imposed on the optimization.
  - Per-node outage constraint.
  - Fixed or minimum data rates per user.



## Appendix: Effect of Normalized Distance



Figure: Optimal fractional in-band power  $\Psi(r)$  as function of normalized transmitter distance  $r = |X_0|/r_{net}$ . Results are shown for three different path-loss coefficients.

Recall: The averaged outage probability is found by

$$F_{\mathsf{Z}}(\Gamma^{-1}) = \int \left(\prod_{i=1}^{M} f_{\Omega_{i}}(\omega_{i})\right) F_{\mathsf{Z}}(\Gamma^{-1}|\boldsymbol{\omega}) d\boldsymbol{\omega}$$
(8)

• In absence of shadowing  $\Omega_i = c_i^{-1} ||X_i||^{-\alpha}$ , which has pdf

$$f_{\Omega_i}(\omega_i) = \begin{cases} \frac{2c_i^{2/\alpha}\omega_i^{-\left(\frac{2+\alpha}{\alpha}\right)}}{\alpha(r_{net}^2 - r_{ex}^2)} & \text{for } c_i r_{net}^{-\alpha} \le \omega_i \le c_i r_{ex}^{-\alpha} \\ 0 & \text{elsewhere} \end{cases}$$
(9)

where  $c_i = (P_i/P_0)$ .

Substituting (9) and (5) into (8), the cdf of Z is found to be

$$\bar{F}_{\mathsf{Z}}(z) = e^{-\beta_0 z} \sum_{s=0}^{m_0 - 1} (\beta_0 z)^s \sum_{t=0}^s \frac{z^{-t}}{(s-t)!} \sum_{\substack{\ell_i \ge 0\\ \sum_{i=0}^M \ell_i = t}} \prod_{i=1}^M \mathbb{E}_{\Omega}(\Omega_i)$$
(10)

where

$$\mathbb{E}_{\Omega}(\Omega_{i}) = p_{\mathsf{n}}\delta_{\ell_{i}} + \frac{2m_{i}^{m_{i}}\Gamma(\ell_{i}+m_{i})}{\alpha(\ell_{i}!)\Gamma(m_{i})\beta_{0}^{(m_{i}+\ell_{i})}} \left\{ p_{\mathsf{c}}\left[I\left(\frac{\psi c_{i}}{r_{\mathsf{net}}^{\alpha}}\right) - I\left(\frac{\psi c_{i}}{r_{\mathsf{ex}}^{\alpha}}\right)\right] + p_{\mathsf{a}}\left[I\left(\frac{K_{s}c_{i}}{r_{\mathsf{net}}^{\alpha}}\right) - I\left(\frac{K_{s}c_{i}}{r_{\mathsf{ex}}^{\alpha}}\right)\right] \right\},$$
(11)

 $c_i = (P_i/P_0)$  and  $_2F_1$  is the Gauss hypergeometric function,

$$I(x) = \frac{{}_{2}F_{1}\left(\left[m_{i}+\ell_{i},m_{i}+\frac{2}{\alpha}\right];m_{i}+\frac{2}{\alpha}+1;-\frac{m_{i}}{x\beta_{0}}\right)}{x^{m_{i}}\left(m_{i}+\frac{2}{\alpha}\right)}$$
(12)

and  $_2F_1([a, b], c, z)$  is the Gauss hypergeometric function.

## Appendix: Average Outage Probability

• In presence of shadowing  $\Omega_i = c_i^{-1} 10^{\xi_i/10} ||X_i||^{-\alpha},$  which has pdf

$$f_{\Omega_{i}}(\omega_{i}) = \begin{cases} \omega^{-\frac{2+\alpha}{\alpha}} \frac{\left[\zeta\left(c_{i}\omega_{i}r_{\mathsf{net}}^{\alpha}\right) - \zeta\left(c_{i}\omega_{i}r_{\mathsf{ex}}^{\alpha}\right)\right]}{\alpha c_{i}^{2/\alpha}\left(r_{\mathsf{net}}^{2} - r_{\mathsf{ex}}^{2}\right)} & \text{for } 0 \leq \omega_{i} \leq \infty \\ 0 & \text{elsewhere} \end{cases}$$

where

$$\zeta(z) = \operatorname{erf}\left(\frac{\sigma_s^2 \ln^2(10) - 50\alpha \ln(z)}{5\sqrt{2}\alpha\sigma_s \ln(10)}\right) e^{\frac{\sigma_s^2 \ln^2(10)}{50\alpha^2}}$$
(13)

Because  $||X_0||$  is deterministic,  $\Omega_0 = 10^{\xi_0/10} ||X_0||^{-\alpha}$  is a log-normal variable with pdf

$$f_{\Omega_0}(\omega) = \frac{10 \left(2\pi\sigma_s^2\right)^{-\frac{1}{2}}}{\ln(10)\omega} \exp\left\{-\frac{10^2 \log_{10}^2 \left(||X_0||^{\alpha}\omega\right)}{2\sigma_s^2}\right\}$$
(14)

for  $0 \leq \omega \leq \infty,$  and zero elsewhere.

Substituting (13), (14) and (5) into (8), the cdf of Z is found to be

$$\bar{F}_{\mathsf{Z}_{M}}(z) = \sum_{s=0}^{m_{0}-1} \sum_{t=0}^{s} \frac{z^{-t}}{(s-t)!} \sum_{\ell_{i} \ge 0} \int_{0}^{\infty} \exp\left\{-\frac{\beta m_{0} z}{\Psi y}\right\} \left(\frac{\beta m_{0} z}{\Psi y}\right)^{s} \sum_{\substack{\Sigma_{i=0}^{M} \ell_{i} = t}}^{m_{0}-1} \sum_{t=0}^{m_{0}-1} \sum_{i=0}^{m_{0}-1} \sum_{t=0}^{m_{0}-1} \sum_{t=0}^{m_{0}-1} \sum_{i=0}^{m_{0}-1} \sum_{i=0}^$$

$$\prod_{i=1}^{M} \left[ p_n \delta_{\ell_i} + p_c \Phi_i(y, \psi) + p_a \Phi_i(y, K_s) \right] f_{\Omega_0}(y) dy$$
(15)

where

$$\Phi_{i}(y,\chi) = \frac{\Gamma(\ell_{i}+m_{i})}{\ell_{i}!\Gamma(m_{i})} \int_{0}^{\infty} f_{\Omega_{i}}(\omega) \left(\frac{\chi\omega}{m_{i}}\right)^{\ell_{i}} \left(\frac{\chi\beta m_{0}\omega}{\psi m_{i}y}+1\right)^{-(m_{i}+\ell_{i})} d\omega.$$
(16)